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## 5. DC-DC 컨버터

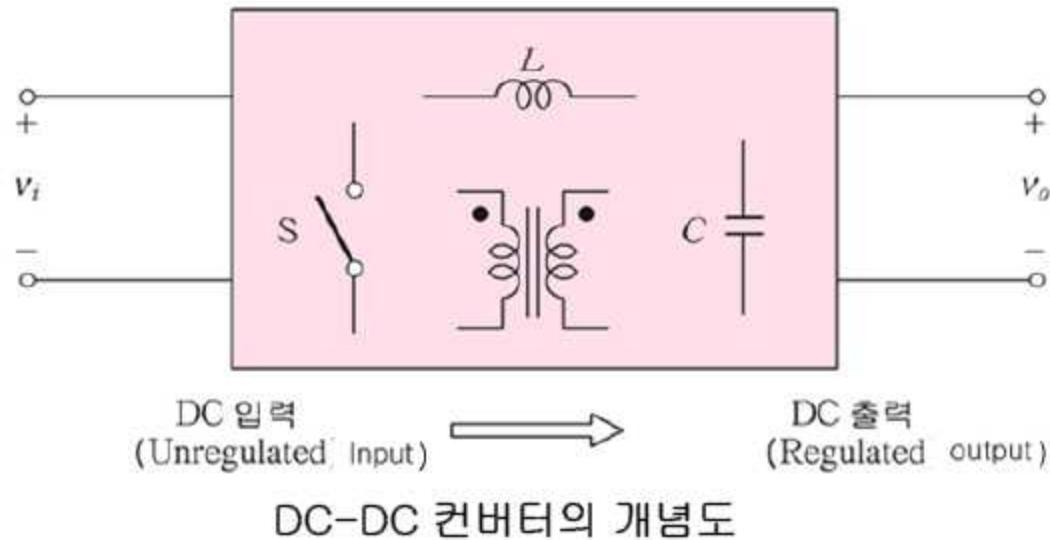


## 제 5 장 DC-DC 컨버터

- 전력변환의 기본원리
- Buck 컨버터
- Boost 컨버터
- Buck-Boost 컨버터
- Forward 컨버터
- Flyback 컨버터



# DC-DC 컨버터의 개념도



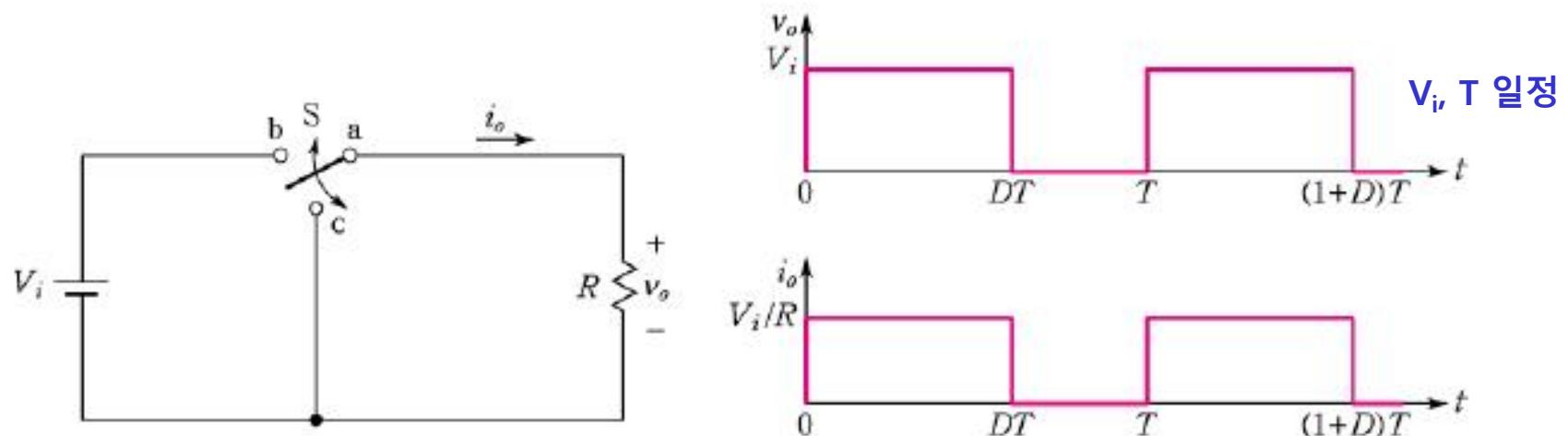
Unregulated dc 입력 → regulated dc 출력

Configuration:

1. Switch (MOSFET, IGBT, Diode) : Energy flow control
2. 인덕터, 커패시터 : 에너지 전달 매개체, Low pass filter (ripple 감소)
3. 변압기 : 전압이득 조절, 전기적인 isolation
4. 제어기 설계



# 간략한 형태의 DC-DC 컨버터



D (duty ratio, 통류율) : 0 ~ 1 범위, 스위치 on 구간

0 ~ DT : switch ab  $\rightarrow v_o = V_i$ , 전력전달구간

DT ~ T : switch ac  $\rightarrow v_o = 0$ , 입력/ 출력 분리

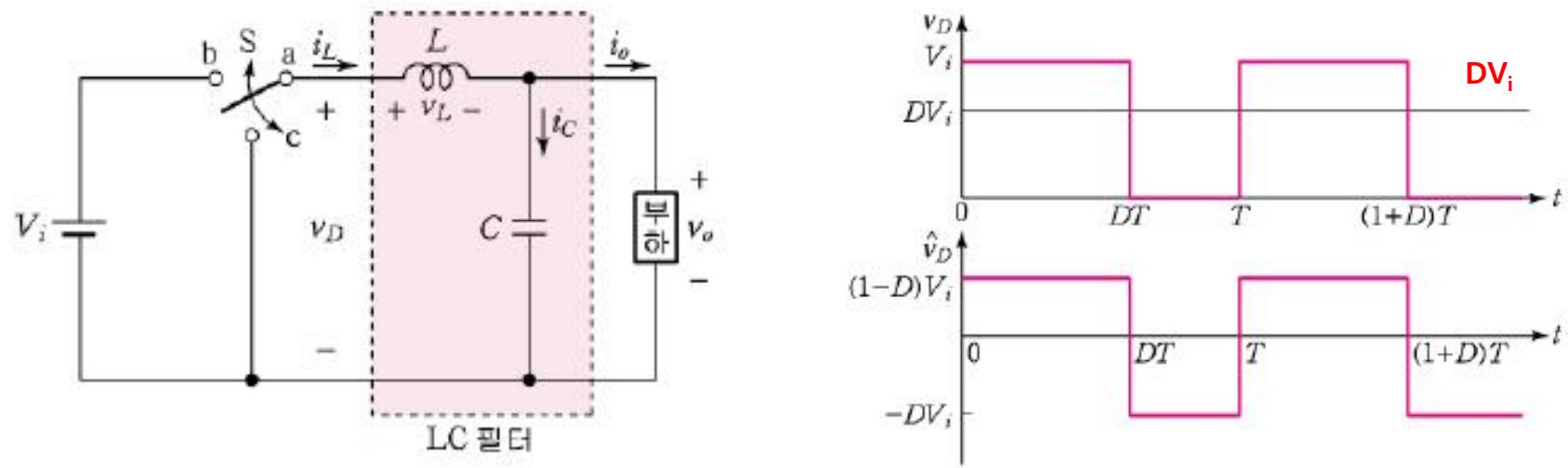
$\langle v_o \rangle, \langle p_o \rangle$  는 D에 비례

$$V_o = \frac{1}{T} \cdot [ DT \cdot V_i + (1-D)T \cdot 0 ] = DV_i$$

$$P_o = \frac{1}{T} \cdot [ DT \cdot \frac{V_i^2}{R} + (1-D)T \cdot 0 ] = D \frac{V_i^2}{R}$$



# Buck 컨버터 기본회로와 파형



$$v_D = v_{DC} + v_{AC} \quad v_{DC} = DV_i$$

$$v_{AC} = v_D - v_{DC} = v_D - DV_i$$

$$G_V \equiv \frac{V_o}{V_i} = \frac{DV_i}{V_i} = D$$

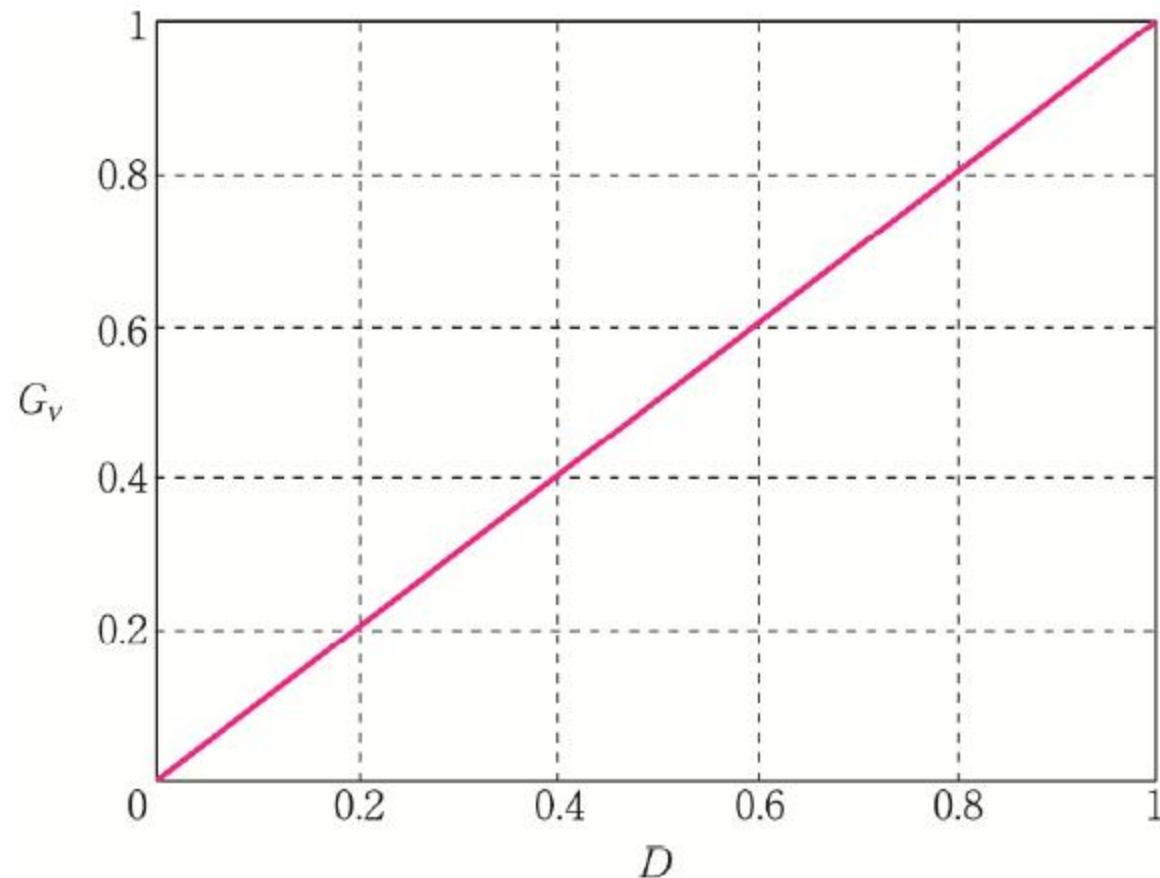
출력전압 : DC+AC 성분

LC low pass filter : 교류성분 차단, regulate 된 직류전압 생성

출력전압  $v_o <$  입력전압  $V_i \rightarrow$  강압형 (Step down)



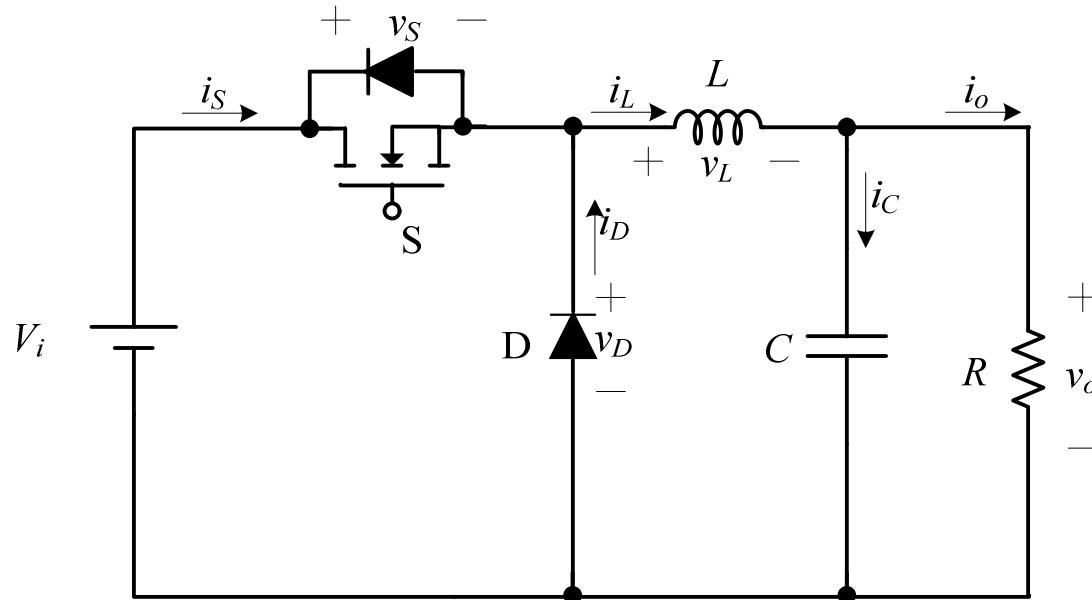
# 통류율(Duty Ratio)과 전압전달비



$$G_V \equiv \frac{V_0}{V_i} = \frac{DV_i}{V_i} = D$$



# Buck 컨버터의 실제적인 구성



MOSFET S : on-off 제어로 입력전원의 에너지를 출력으로 전달 (active switch)  
Diode D : MOSFET가 off 되는 구간 동안 인덕터 전류  $i_L$ 이 흐를수 있도록 path 제공

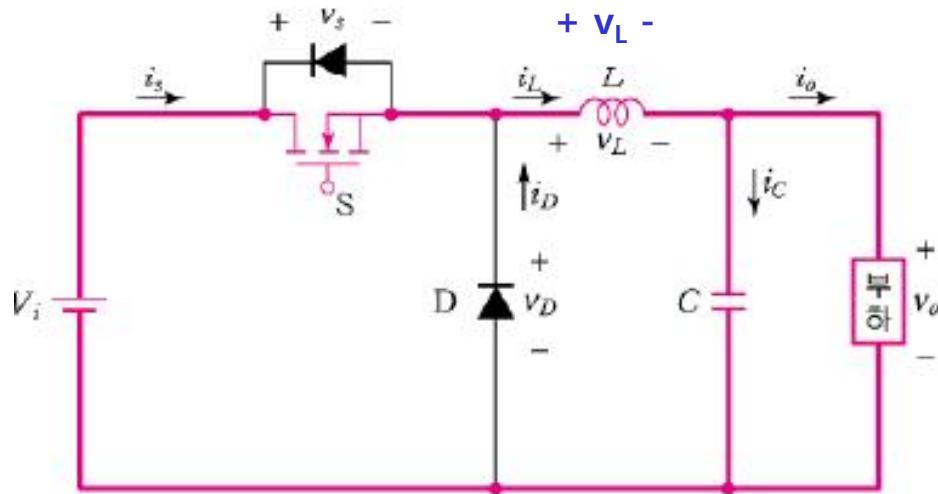
→ freewheeling path (passive switch)

$0 < t < DT$  : S on,  $v_D = V_i$  → reverse bias → off

$DT < t < T$  : S off,  $v_D = 0$  → forward bias → on



# DT 구간 : 스위치 on



S on :  $v_S = 0, i_S = i_L$   
D off :  $v_D = V_i, i_D = 0$   
 $v_L = V_i - V_o > 0$  ( $V_i > V_o$ )  
→  $i_L$  증가 (상승)  
 $P_L = v_L i_L > 0$   
 $E_L = \frac{1}{2} L i_L^2 \rightarrow$  증가

$0 < t < DT$  (MOSFET S on, Diode D off)

$$< v_o > < V_i$$

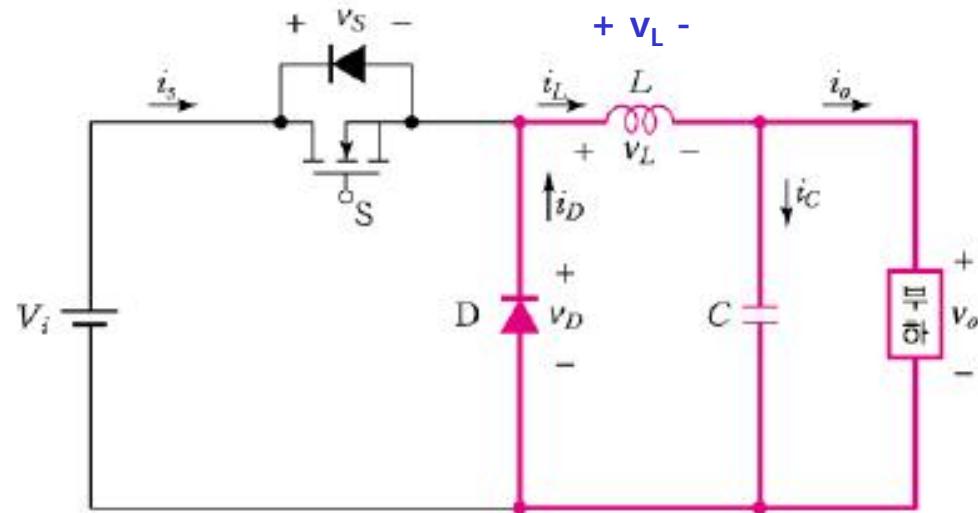
정상상태 : 한 주기의 시작과 끝에서  $i_L, v_o$ 의 값이 같다.

- 인덕터 전류  $i_L$ , 커패시터 전압  $v_o$  = 출력전압

L & C are finite



# (1-D)T 구간 : Diode on

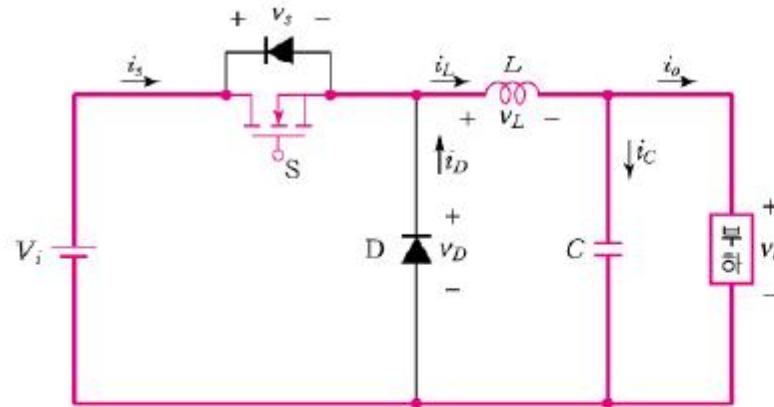


S off :  $v_S = V_i$ ,  $i_S = 0$   
D on :  $v_D = 0$ ,  $i_D = i_L$   
 $v_L = -V_o < 0$   
 $\rightarrow i_L$  감소 (하강)  
 $P_L = v_L i_L < 0$   
 $E_L = \frac{1}{2}L i_L^2 \rightarrow$  감소

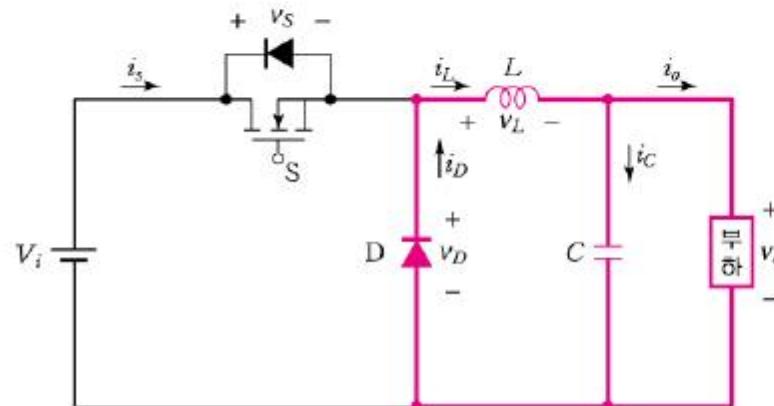
$DT < t < T$  (MOSFET S off, Diode D on)



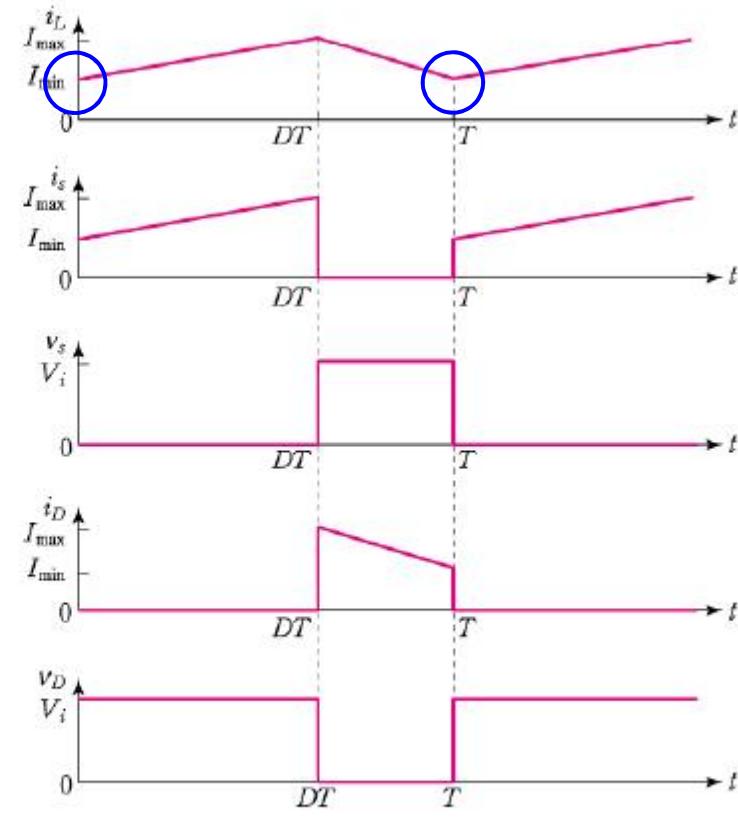
# Buck 컨버터 회로 각부의 파형



$0 < t < DT$  (MOSFET S on, Diode D off)



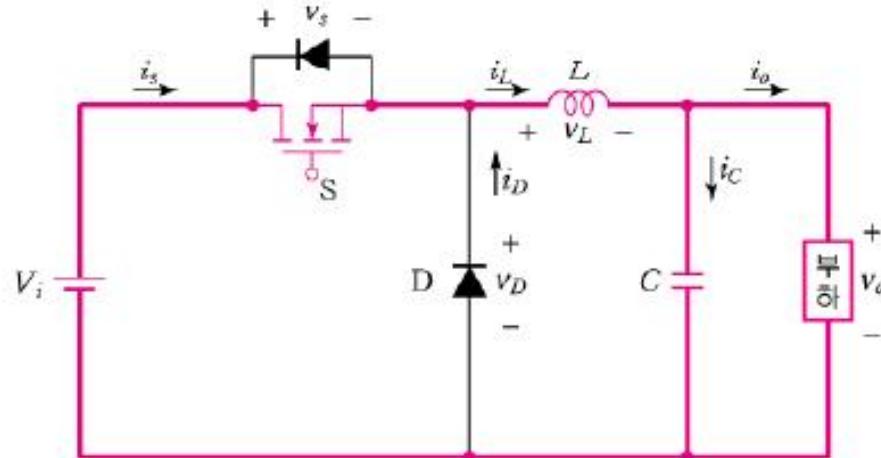
$DT < t < T$  (MOSFET S off, Diode D on)



S: ON, D: OFF    S: OFF, D: ON    S: ON, D: OFF



# 인덕터 전류 : 전류상승구간 ( $0 < t < DT$ )

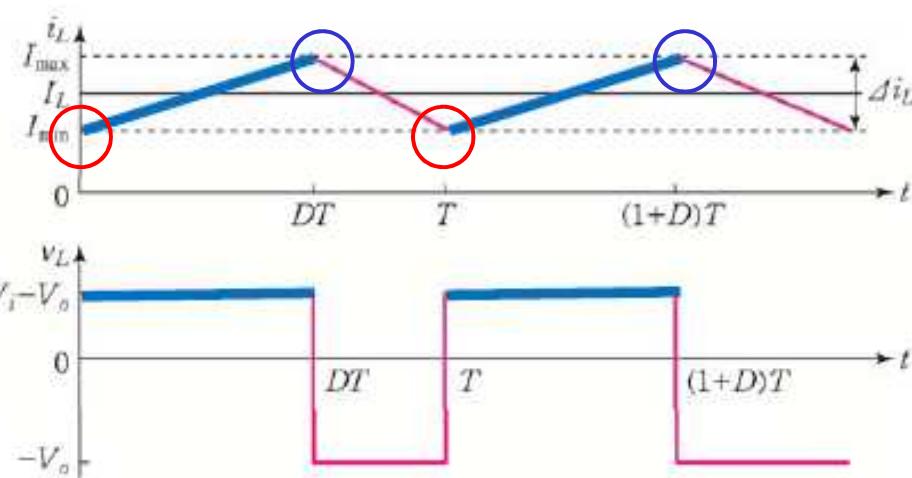


i)  $0 \leq t < DT$

$$v_L = V_i - V_0$$

$$v_L = L \frac{di_L}{dt}$$

$$\frac{di_L}{dt} = \frac{V_i - V_0}{L} \quad i_L(t) = \frac{1}{L} \int_0^t (V_i - V_0) dt + I_{\min}$$



DT 구간 동안 인덕터 전류의 상승폭  $\Delta i_L$

$$\Delta i_L = \frac{V_i - V_0}{L} \cdot DT$$

$$I_{\max} = I_L + \frac{1}{2} \Delta i_L = I_o + \frac{V_i - V_0}{2L} \cdot DT$$

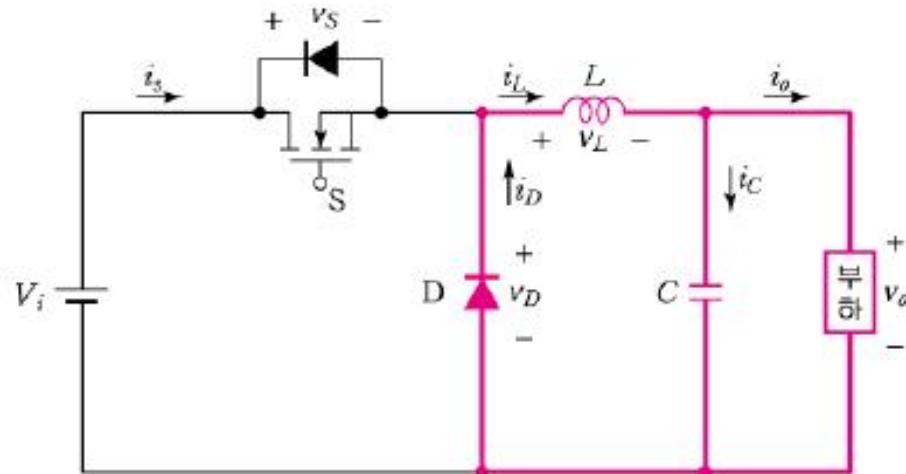
$$= I_o + \frac{V_i}{2L} \cdot (1 - D) \cdot DT$$

$$I_{\min} = I_L - \frac{1}{2} \Delta i_L = I_o - \frac{V_i - V_0}{2L} \cdot DT$$

$$= I_o - \frac{V_i}{2L} \cdot (1 - D) \cdot DT$$



# 인덕터 전류 : 전류하강구간 ( $DT < t < T$ )



ii)  $DT \leq t < T$

$$v_L = -v_0$$

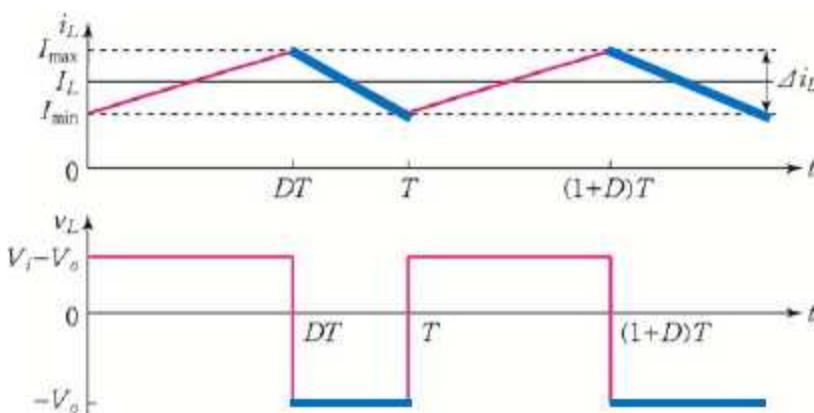
$$\frac{di_L}{dt} = -\frac{v_0}{L} \quad i_L(t) = \frac{1}{L} \int_{DT}^t (-V_0) dt + I_{\max}$$

(1-D)T 구간 동안 인덕터 전류의 하강폭  $\Delta i_L$

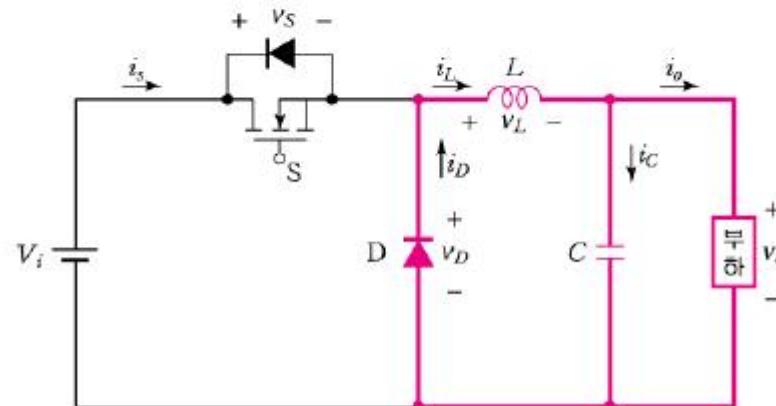
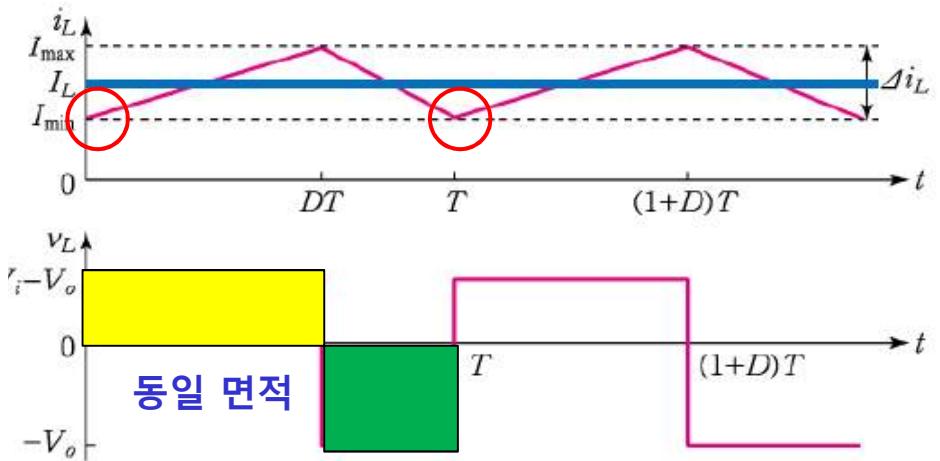
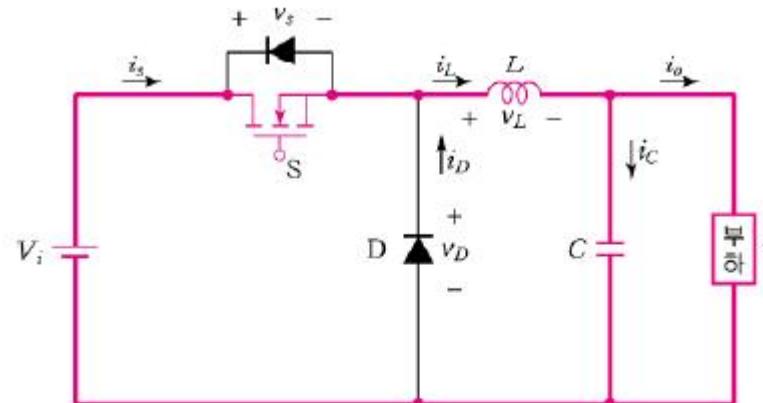
$$\Delta i_L = \frac{V_0}{L} \cdot (1-D)T$$

$$I_{\max} = I_L + \frac{1}{2} \Delta i_L = I_o + \frac{V_0}{2L} \cdot (1-D)T$$

$$I_{\min} = I_L - \frac{1}{2} \Delta i_L = I_o - \frac{V_0}{2L} \cdot (1-D)T$$



# 인덕터 전압파형



정상상태에서 상승폭  $\Delta i_L$  = 하강폭  $\Delta i_L$

$$\frac{V_i - V_o}{L} DT = \frac{V_o}{L} (1-D)T$$

$$(V_i - V_o)DT = V_o(1-D)T$$

$$V_o = DV_i$$

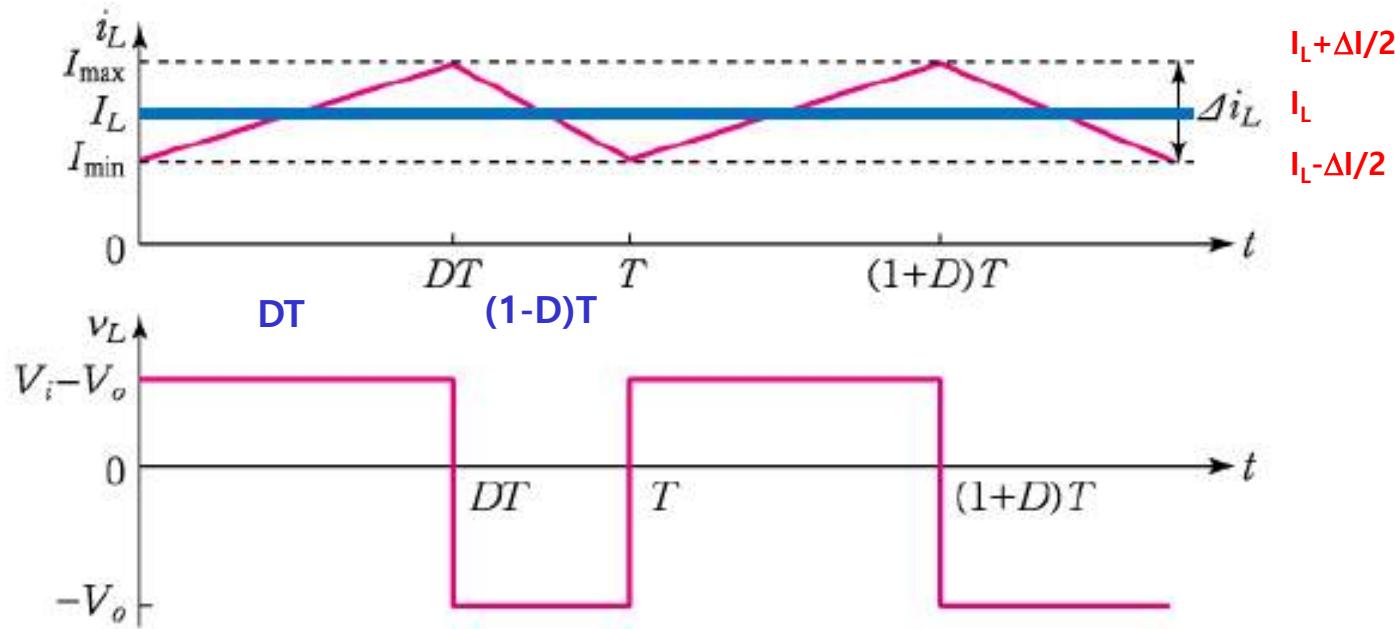
인덕터 전압 x 시간은 DT와 (1-D)T 구간에서 동일

$$(V_i - V_o) \cdot DT = V_o \cdot (1-D)T$$

$$V_o = DV_i$$



# 인덕터 평균전류 $i_L$



$$i_L = i_C + i_o$$

정상상태에서 커패시터 전류의 평균값  $I_C = 0$ 이므로

$$I_L = I_o$$

$$i_L = I_L \text{ (직류성분)} + \hat{i}_L \text{ (교류성분)}$$

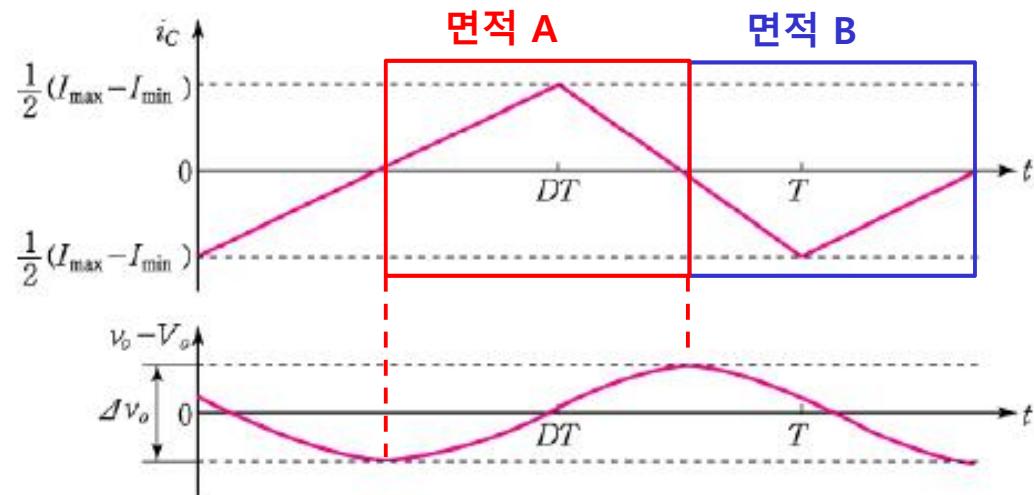
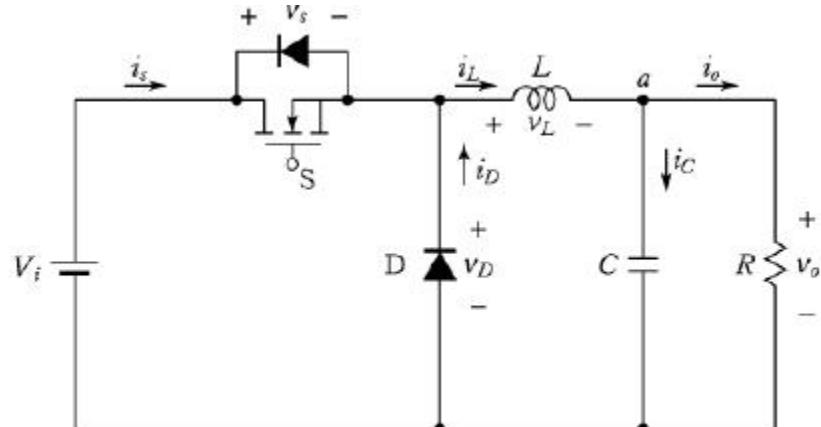
인덕터 전류 :  $v_L/L$ 의 기울기로

$DT$  (S on, D off) 구간 동안 선형적으로 증가

$(1-D)T$  (S off, D on) 구간 동안 선형적으로 감소



# 출력 전압 $v_o$



$$I_0 = \frac{V_0}{R}$$

$$i_C = i_L - I_o = C \frac{dv_o}{dt}$$

$$\Delta v_0 = \frac{1}{C} \frac{1}{2} \left( \frac{I_{\max} - I_{\min}}{2} \right) \bullet \frac{T}{2}$$

$$= \frac{1}{C} (I_{\max} - I_{\min}) \bullet \frac{T}{8}$$

$$I_{\max} - I_{\min} = \frac{V_i - DV_i}{L} DT$$

$$\Delta v_0 = \frac{1}{LC} \frac{V_i(1-D)D \bullet T^2}{8}$$

면적 A

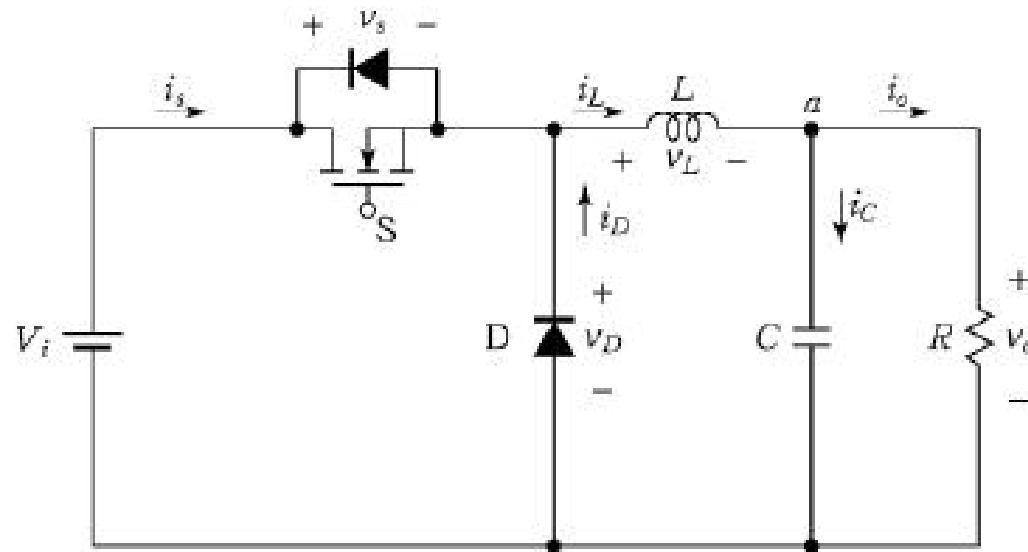


## 예제 5-2

<예제 5-2>

$$V_i = 100, f_s = 100\text{KH}, D = 0.6, R = 10, L = 50\mu\text{H}, C = 100\mu\text{F}$$

- (a)  $V_o, I_o$  (b)  $I_{\max}, I_{\min}$  (c)  $\Delta v_o$



DC transformer

$$P_{in} = V_i I_s$$

$$P_o = V_o I_o = V_o I_L$$

$$P_{in} = P_o \quad V_o = D V_i$$

$$I_s = \frac{V_o}{V_i} I_o = D I_o$$



## 예제 5-2

입력전압 100V, 100 kHz, D=0.6, R =10  
ohm, L = 10 uH, C=100 uF

<풀이>

$$V_0 = DV_i = 0.6 \times 100 = 60 \dots V$$

$$I_L = I_0 = \frac{V_0}{R} = \frac{60}{10} = 6 \quad A$$

$$\begin{aligned}\Delta v_0 &= \frac{1}{LC} \frac{V_i(1-D)D \bullet T^2}{8} \\ &= \frac{1}{100 \times 10^{-6}} \times (8.4 - 3.6) \times \frac{10^{-5}}{8} = \frac{4.8}{80} = 0.06 \quad V\end{aligned}$$

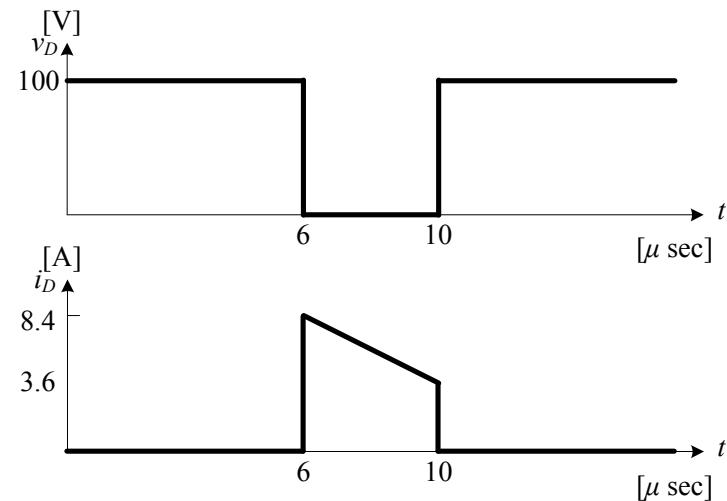
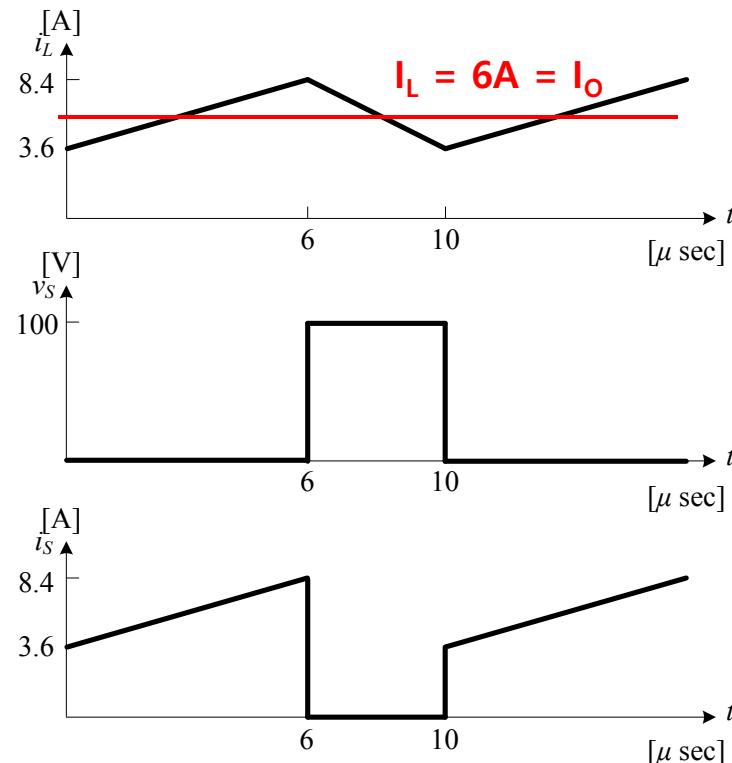
$$\begin{aligned}I_{\max} &= I_L + V_0 \bullet \frac{1-D}{2L} \bullet T \\ &= 6 + 60 \times \frac{1-0.6}{2 \times 50 \times 10^{-6}} \times 10^{-5} = 6 + 2.4 = 8.4 \quad A\end{aligned}$$

$$\begin{aligned}I_{\max} &= I_L - V_0 \bullet \frac{1-D}{2L} \bullet T \\ &= 6 - 60 \times \frac{1-0.6}{2 \times 50 \times 10^{-6}} \times 10^{-5} = 6 - 2.4 = 3.6 \quad A\end{aligned}$$



## 예제 5-3

<예제 5-3> 예제 5-2에서  $i_L$ ,  $v_S$ ,  $i_S$ ,  $v_0$ ,  $i_D$  파형



# 전류 불연속 모드

## 전류 불연속 모드 (Discontinuous mode) 조건

$$I_{\min} = I_L - V_o \frac{1-D}{2L} T = I_o - V_i \frac{D(1-D)}{2L} T < 0$$

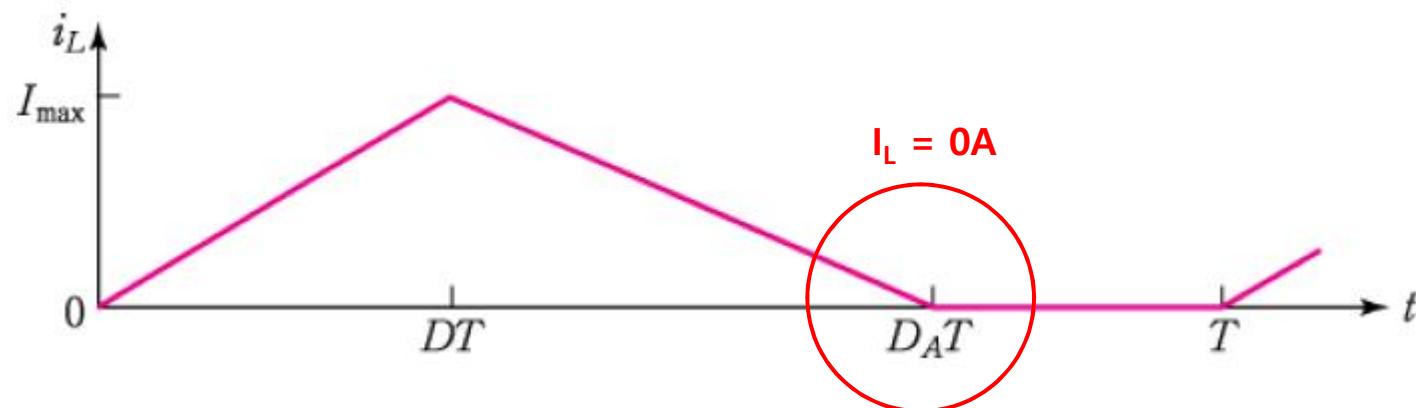
1) 인덕터전류  $i_L$

$$I_{\min} = 0$$

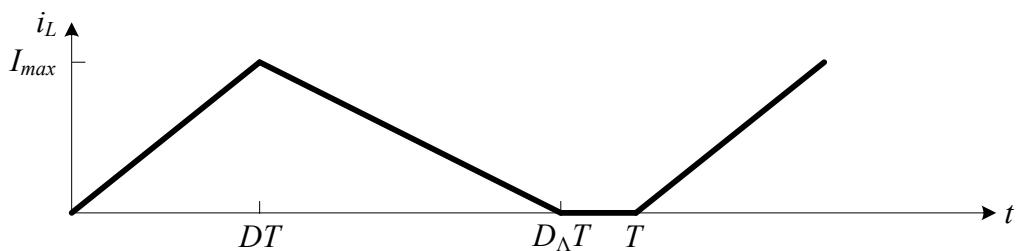
$D_A T < t < T : i_L = 0 \rightarrow$  switch S와 diode D가 동시에 off

2) 출력전압  $V_o$

컨버터 입력에너지  $E_{in}$  = 출력으로 유출되는 에너지  $E_o$



# 불연속 모드 정상상태 해석



(i)  $0 \leq t \leq DT$

$$i = \frac{V_i - V_0}{L} \cdot t$$

$$I_{\max} = \frac{V_i - V_0}{L} \cdot DT$$

(ii)  $DT \leq t \leq D_A T$

$$i = I_{\max} - \frac{V_0}{L} (t - DT)$$

(iii)  $D_A T \leq t \leq T$

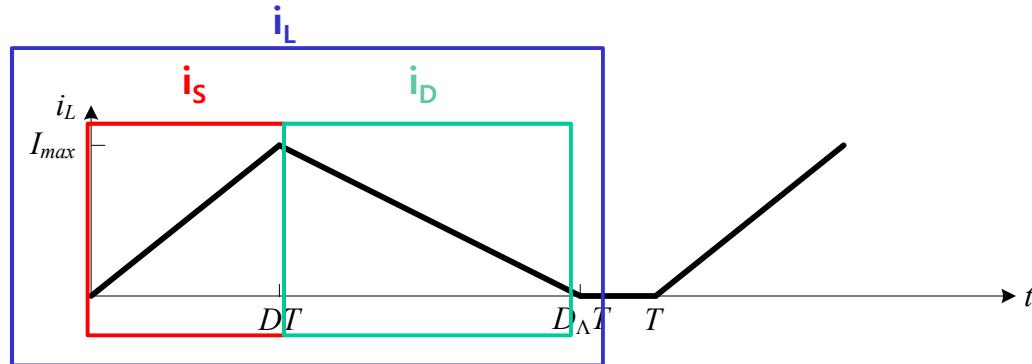
$$i = 0$$

$$0 = I_{\max} - \frac{V_0}{L} (D_A T - DT) \Rightarrow$$

$$D_A = D + \frac{L \cdot I_{\max}}{V_0 T}$$



# 불연속 모드 정상상태 해석



컨버터 입력 에너지  $E_{in}$  = 출력에너지  $E_0$ 는 같으므로

$$E_{in} = V_i \cdot \frac{I_{max}}{2} \cdot DT$$

$$E_0 = V_0 \cdot \frac{I_{max}}{2} \cdot D_A T$$

$$G_V \equiv \frac{V_0}{V_i} = \frac{D}{D_A}$$

연속전류 :  $D_A = 1 \rightarrow G_V = D$   
불연속전류  $\rightarrow$  출력전압 증가

$$i_L(t = D_A T) = \frac{1}{L} \int_{DT}^{D_A T} (-V_0) dt + I_{max} = 0$$

$$I_{max} = \frac{V_0}{L} (D_A - A) \cdot T$$

$$D_A = D + \frac{LI_{max}}{TV_0}$$

출력전류 평균  $I_o$  = 인덕터 전류 평균  $i_L$

$$I_L = I_o = \frac{I_{max}}{2} D_A \Rightarrow I_{max} = \frac{2I_o}{D_A}$$

$$D_A = D + \frac{2LI_0}{D_A TV_0} = D + \frac{2LI_0}{DTV_i}$$