

Fourier Series

□ 주기 [전력] 신호인 경우, 기본 주기 T_0

□ 정현 Fourier Series

- 직교 함수 집합 : 구간 $t_0 \leq t \leq t_0 + T_0$

$$\{\phi_n(t)\} = \{1, \cos n\omega_0 t, \sin n\omega_0 t, n=1,2,3,\dots,\infty\}$$

□ 복소 지수함수 Fourier Series

- 직교 함수 집합 : 구간 $t_0 \leq t \leq t_0 + T_0$

$$\{\phi_n(t)\} = \{e^{jn\omega_0 t}, n=0, \pm 1, \pm 2, \dots, \pm\infty\} \left(\omega_0 = \frac{2\pi}{T_0} \right)$$

복소 지수함수 Fourier Series

□ 복소 지수함수 근저 함수 (basis function) 집합에 의한 표현

- 구간 $t_0 \leq t \leq t_0 + T_0$ 에서 임의의 신호 $x(t)$ 는 아래 선형 조합으로 표현

$$x(t) = \sum_{n=-\infty}^{\infty} C_n e^{jn\omega_0 t}, \quad \omega_0 = \frac{2\pi}{T_0}$$

- 선형조합의 계수 C_n 은 다음과 같이 구함

$$C_n = \frac{\langle x(t), \phi_n(t) \rangle}{\int_{t_0}^{t_0+T_0} |\phi_n(t)|^2 dt} = \frac{\int_{t_0}^{t_0+T_0} x(t) \phi_n^*(t) dt}{\int_{t_0}^{t_0+T_0} 1 dt} = \frac{1}{T_0} \int_{t_0}^{t_0+T_0} x(t) e^{-jn\omega_0 t} dt$$

- 만일 신호 $x(t)$ 가 주기 T_0 를 가진 주기 신호라면 basis 함수들도 같은 주기의 주기 함수가 되므로 위의 무한급수 표현은 한 주기의 구간을 넘어서 모든 시구간 $-\infty < t < \infty$ 로 확장할 수 있음

(3) The inner product in time :

$$\text{Let's define : } \langle x(t), e^{jk\omega_0 t} \rangle \equiv \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt = X(k)$$

$$\text{then : } \langle e^{jm\omega_0 t}, e^{jk\omega_0 t} \rangle = \frac{1}{T} \int_0^T e^{jm\omega_0 t} e^{-jk\omega_0 t} dt = \frac{1}{T} \int_0^T e^{j(m-k)\omega_0 t} dt$$

$$\begin{aligned} [\because \omega_0 T = 2\pi] &= \begin{cases} \frac{1}{T} \int_0^T 1 dt = 1, & m = k \\ \frac{1}{T} \left[\frac{e^{j(m-k)\omega_0 t}}{j(m-k)\omega_0} \right]_0^T = 0, & m \neq k \end{cases} \\ &= \delta(m-k) \end{aligned}$$

- Fourier Series 및 Fourier Series 계수

$$x(t) = \sum_{n=-\infty}^{\infty} C_n e^{jn\omega_0 t} = \sum_{n=-\infty}^{\infty} C_n e^{j2\pi n t/T_0} \quad (\text{Fourier Series: Synthesis})$$

$$C_n = \frac{1}{T_0} \int_{T_0} x(t) e^{-jn\omega_0 t} dt = \frac{1}{T_0} \int_{T_0} x(t) e^{-j2\pi n t/T_0} dt \quad (\text{FS Coefficients: Analysis})$$

- FA Coefficient C_n 은 $n\omega_0$ 주파수 성분(n th harmonic)의 전력량
- 일명, 스펙트럼 계수(spectrum coefficient)
- 복소수 : $C_n = |C_n| e^{j \arg C_n}$, 여기서 $\arg C_n$ 은 C_n 의 편각(argument)

Fourier Series 성질

□ 대칭성

- $x(t)$ 의 값이 real인 경우, 즉 $x(t) = x^*(t)$ 인 경우

$$c_n = c_{-n}^* \quad (\text{conjugate symmetric})$$

$$\Rightarrow |c_{-n}| = |c_n| \quad (\text{amplitude spectrum is even})$$

$$\arg c_{-n} = -\arg c_n \quad (\text{phase spectrum is odd})$$

- If $x(t)$ is real & even (i.e. $x(t) = x(-t)$), then c_n is also real & even
- If $x(t)$ is real & odd (i.e. $x(t) = -x(-t)$), then c_n is imaginary & odd

□ 선형성

- $x(t)$ 와 $y(t)$ 가 동일한 주기를 갖는 주기신호인 경우

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

$$y(t) = \sum_{n=-\infty}^{\infty} d_n e^{jn\omega_0 t}$$

- $z(t) = \alpha x(t) + \beta y(t)$ 의 푸리에 급수는

$$z(t) = \sum_{n=-\infty}^{\infty} g_n e^{jn\omega_0 t}$$

$$\text{with } g_n = \alpha c_n + \beta d_n$$

□ 시간 천이

- $x(t)$ 의 푸리에 계수가 c_n 인 경우
- $x(t-\tau)$ 의 푸리에 계수 d_n 은

$$\begin{aligned}d_n &= \frac{1}{T_0} \int_{T_0} x(t-\tau) e^{-jn\omega_0 t} dt = \frac{1}{T_0} \int_{T_0} x(t-\tau) e^{-jn\omega_0(t-\tau)} e^{-jn\omega_0\tau} dt \\&= e^{-jn\omega_0\tau} \frac{1}{T_0} \int_{T_0} x(\lambda) e^{-jn\omega_0\lambda} d\lambda \\&= e^{-jn\omega_0\tau} \cdot c_n\end{aligned}$$

- 진폭 스펙트럼은 변화가 없다.

$$|d_n| = |c_n|$$

Parseval's Theorem

□ 주기 신호의 평균 전력

$$P = \frac{1}{T_0} \int_{T_0} |x(t)|^2 dt = \sum_{n=-\infty}^{\infty} |c_n|^2$$

- Real valued signal의 경우

$$P = |c_0|^2 + 2 \sum_{n=1}^{\infty} |c_n|^2$$

- $|c_n|^2$ is called power spectrum

□ [proof]

- $z(t) = x(t)y^*(t)$ 의 푸리에 계수

$$\begin{aligned} z(t) = x(t)y^*(t) &= \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t} \sum_{m=-\infty}^{\infty} d_m^* e^{-jm\omega_0 t} \\ &= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} c_n d_m^* e^{j(n-m)\omega_0 t} = \sum_{k=-\infty}^{\infty} \left(\sum_{m=-\infty}^{\infty} c_{k+m} d_m^* \right) e^{jk\omega_0 t} \end{aligned}$$

- $z(t)$ 의 푸리에 계수

$$\sum_{m=-\infty}^{\infty} c_{k+m} d_m^* = \frac{1}{T_0} \int_{T_0} x(t)y^*(t) e^{-jk\omega_0 t} dt$$

- $z(t)$ 의 직류값($k=0$ 에서의 푸리에 계수)

$$\sum_{m=-\infty}^{\infty} c_m d_m^* = \frac{1}{T_0} \int_{T_0} x(t)y^*(t) dt$$

- $x(t) = y(t)$ 인 경우

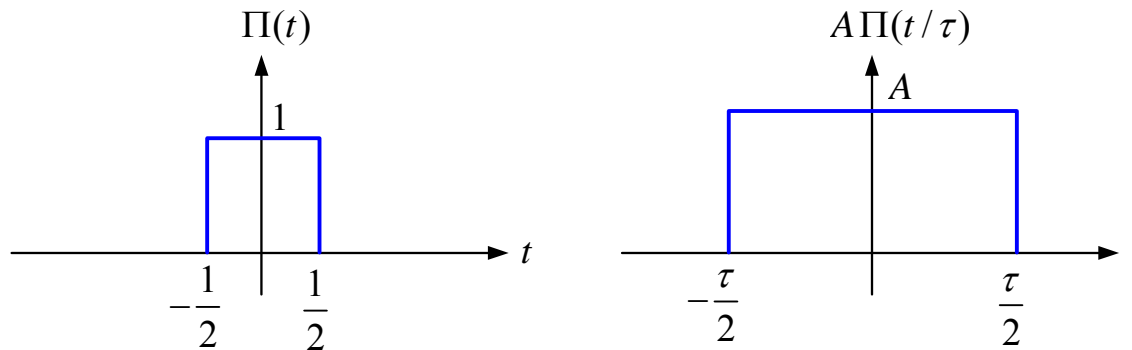
$$\sum_{n=-\infty}^{\infty} |c_n|^2 = \frac{1}{T_0} \int_{T_0} |x(t)|^2 dt$$

Rectangular Pulse

- 크기가 1이고 펄스 폭이 1인 사각 펄스(구형파)

$$\begin{aligned}\Pi(t)[= \text{rect}(t)] &= \begin{cases} 1 & \text{for } -\frac{1}{2} < t < \frac{1}{2} \\ 0 & \text{otherwise} \end{cases} \\ &= u\left(t + \frac{1}{2}\right) - u\left(t - \frac{1}{2}\right)\end{aligned}$$

- 크기가 A이고 펄스 폭이 τ 인 사각 펄스: $A\Pi(t/\tau)$

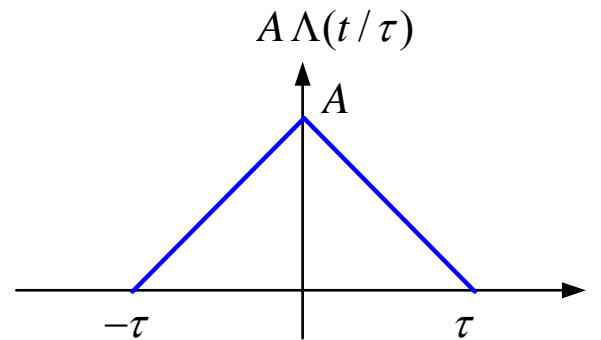
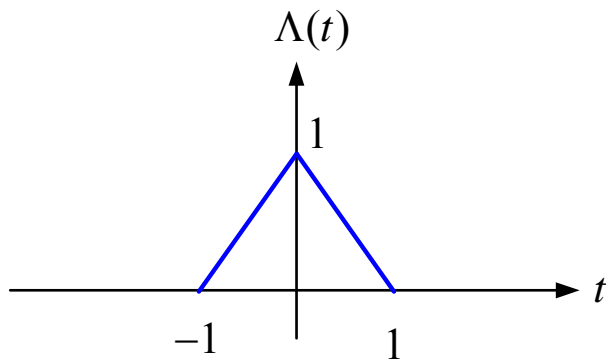


Triangular Pulse

- 크기가 1이고 펄스 폭이 2인 삼각 펄스

$$\Lambda(t) = \begin{cases} 1 - |t| & \text{for } -1 < t < 1 \\ 0 & \text{otherwise} \end{cases}$$

- 크기가 A 이고 펄스 폭이 2τ 인 삼각 펄스: $A\Lambda(t/\tau)$

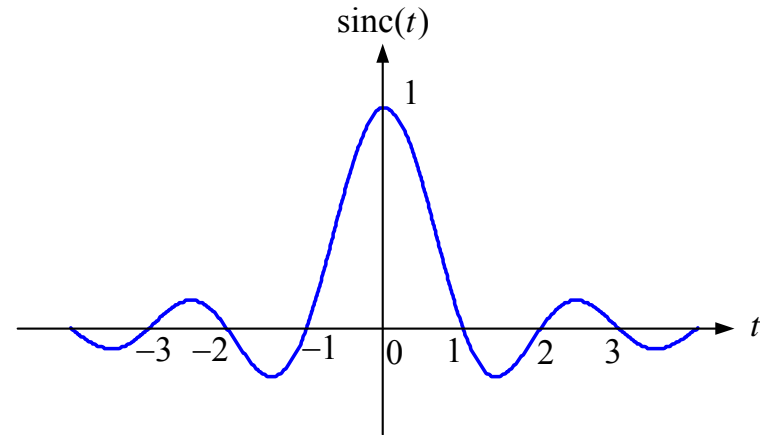
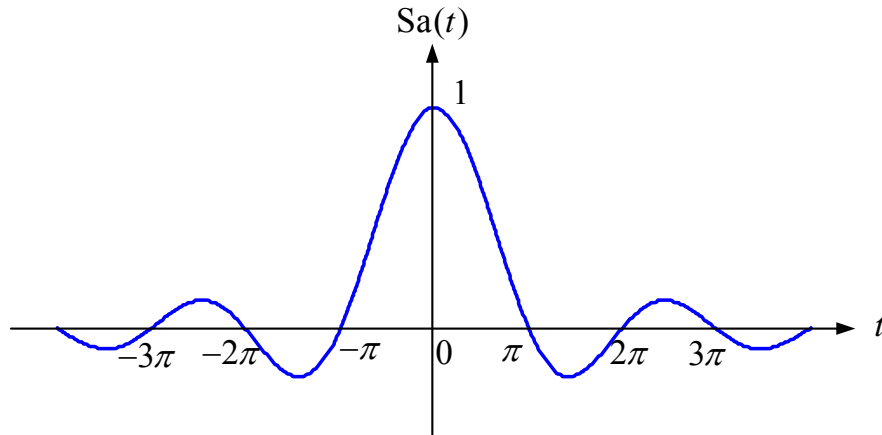


Sampling Function

□ Sa 함수와 sinc 함수

$$\text{Sa}(t) = \frac{\sin t}{t}$$

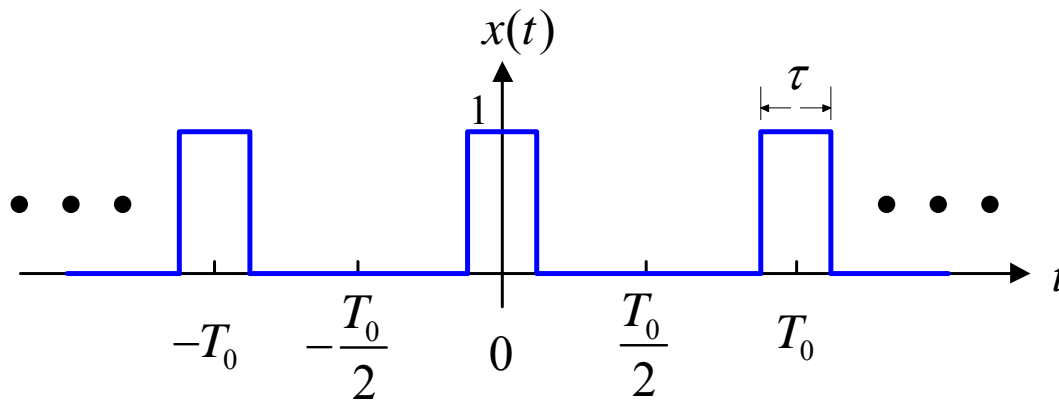
$$\text{sinc}(t) = \frac{\sin \pi t}{\pi t} = \text{Sa}(\pi t)$$



□ [Example] Rectangular Pulse Train

- 주기가 T_0 이고 펄스폭이 τ 인 구형 펄스 열

$$x(t) = \sum_{k=-\infty}^{\infty} \Pi\left(\frac{t-kT_0}{\tau}\right)$$



□ [Example] Rectangular Pulse Train

$$\omega_0 = 2\pi / T_0 = \pi / 2$$

$$\begin{aligned} c_n &= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) e^{-jn\omega_0 t} dt = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) \exp\left(\frac{-j2\pi nt}{T_0}\right) dt \\ &= \frac{1}{T_0} \int_{-\tau/2}^{\tau/2} (1) \exp\left(\frac{-j2\pi nt}{T_0}\right) dt \end{aligned}$$

i) $n = 0$: dc value

$$c_0 = \frac{1}{T_0} \int_{-\tau/2}^{\tau/2} (1) dt = \frac{\tau}{T_0}$$

ii) $n \neq 0$

$$\begin{aligned}c_n &= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) \exp\left(\frac{-j2\pi nt}{T_0}\right) dt = \frac{1}{T_0} \int_{-\tau/2}^{\tau/2} (1) \exp\left(\frac{-j2\pi nt}{T_0}\right) dt \\&= \frac{1}{-j2\pi n} \exp\left(\frac{-j2\pi nt}{T_0}\right) \Big|_{-\tau/2}^{\tau/2} \\&= \frac{1}{j2\pi n} \left[\exp\left(\frac{j\pi n\tau}{T_0}\right) - \exp\left(\frac{-j\pi n\tau}{T_0}\right) \right] \\&= \frac{\sin(\pi n\tau / T_0)}{\pi n} = \frac{\tau}{T_0} \frac{\sin(\pi n\tau / T_0)}{(\pi n\tau / T_0)} \\&= \frac{\tau}{T_0} \operatorname{sinc}\left(\frac{n\tau}{T_0}\right) = \frac{\tau}{T_0} \operatorname{sinc}(nf_0\tau) \\&= \frac{\tau}{T_0} \operatorname{Sa}\left(\frac{\pi n\tau}{T_0}\right) = \frac{\tau}{T_0} \operatorname{Sa}\left(\frac{n\omega_0\tau}{2}\right)\end{aligned}$$

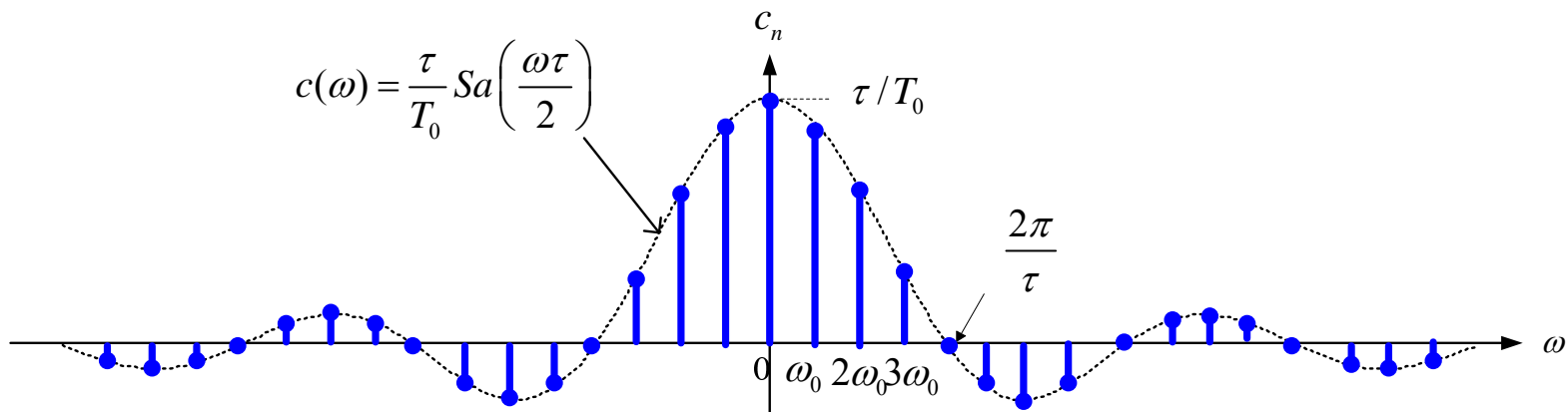
□ [Example] Rectangular Pulse Train

- Spectrum shape :

$$c_n = c(n\omega_0) = \frac{\tau}{T_0} \text{Sa}\left(\frac{n\omega_0\tau}{2}\right) = \frac{\tau}{T_0} \text{Sa}\left(\frac{\omega\tau}{2}\right) \Big|_{n\omega_0}$$

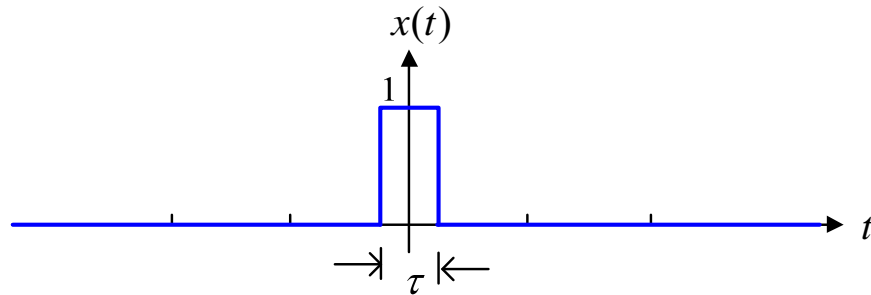
$$c(\omega) = \frac{\tau}{T_0} \text{Sa}\left(\frac{\omega\tau}{2}\right)$$

- $c(\omega)$ 의 파형은 진동하면서 감소하고, $\omega = n\pi / \tau$ ($n \neq 0$)마다 zero-crossing
- First zero-crossing frequency: $\omega = \pi / \tau$

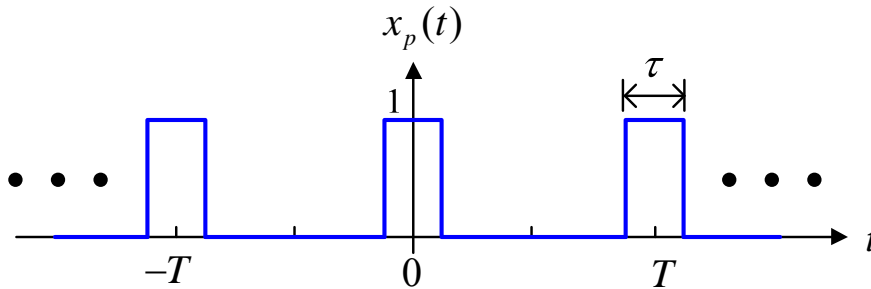


비주기 신호의 Fourier Transform

□ 비주기 신호의 주기적 확장



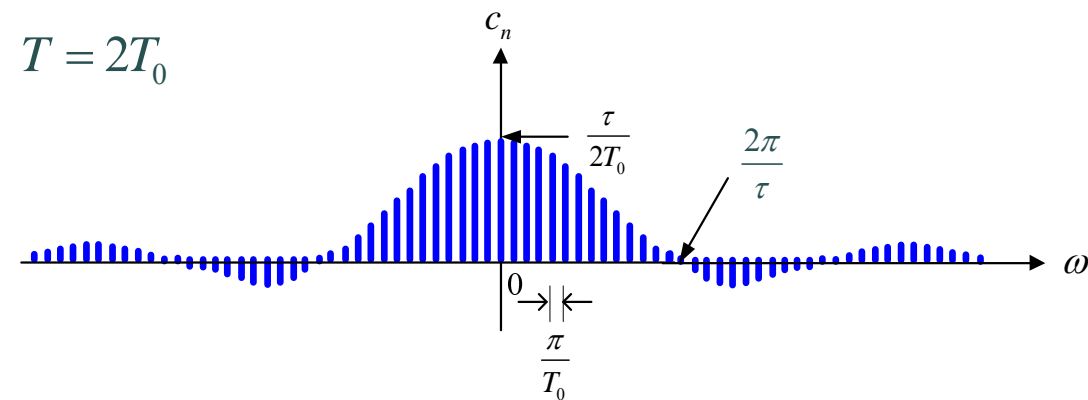
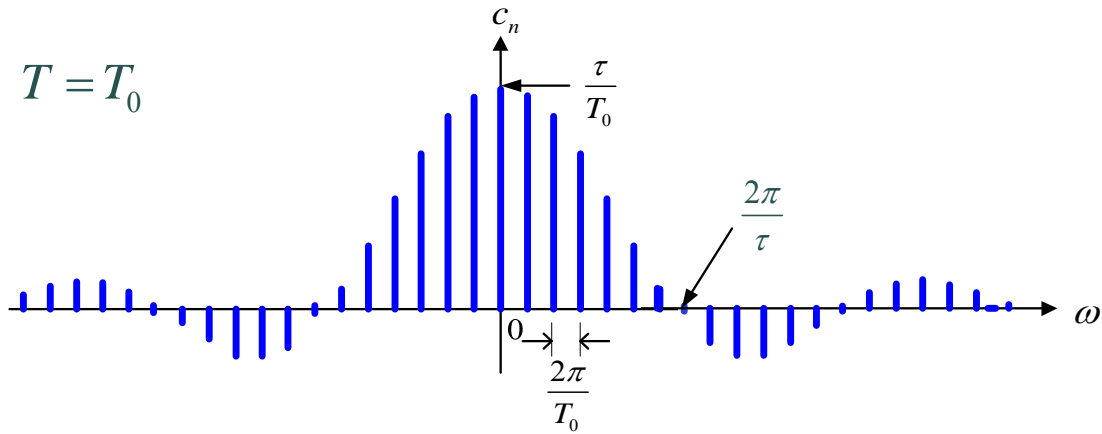
$T \rightarrow \infty$ ↑ ↓ periodic extension



$$x_p(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t}$$

$$c_n = \frac{1}{T} \int_{-T/2}^{T/2} x_p(t) e^{-jn\omega_0 t} dt \quad (\omega_0 = 2\pi / T)$$

$$c_n = \frac{\tau}{T} \text{Sa} \left(\frac{\omega\tau}{2} \right) \Big|_{\omega=n\omega_0} = \frac{\tau}{T} \text{Sa} \left(\frac{n\omega_0\tau}{2} \right)$$



1. Spectrum shape :

$$c(\omega) = \frac{\tau}{T} \text{Sa}\left(\frac{\omega\tau}{2}\right)$$

2. First zero-crossing : $\frac{\omega\tau}{2} = \pi$

$$\omega = n\omega_0 = n\frac{2\pi}{T} = \frac{2\pi}{\tau}$$

□ Define discrete frequency ω_n as

$$\omega_n \triangleq n\omega_0 = n \frac{2\pi}{T}$$

□ Define $X(\omega_n)$

$$X(\omega_n) \triangleq Tc_n = \int_{-T/2}^{T/2} x_p(t) e^{-jn\omega_0 t} dt$$

□ Let $T \rightarrow \infty$ then

$$X(\omega_n) = Tc_n \xrightarrow{T \rightarrow \infty} \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \triangleq X(\omega)$$

$$\begin{aligned} x(t) &= \lim_{T \rightarrow \infty} x_p(t) = \lim_{T \rightarrow \infty} \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t} = \lim_{T \rightarrow \infty} \frac{1}{T} \sum_{n=-\infty}^{\infty} Tc_n e^{jn\omega_0 t} \\ &= \lim_{\substack{T \rightarrow \infty \\ \omega_0 \rightarrow 0}} \sum_{n=-\infty}^{\infty} [Tc_n] e^{jn\omega_0 t} \cdot \frac{\omega_0}{2\pi} = \int_{-\infty}^{\infty} [X(\omega)] e^{j\omega t} \frac{d\omega}{2\pi} \end{aligned}$$

□ Fourier Transform

$$X(\omega) = F \{x(t)\} = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

□ Inverse Fourier Transform

$$x(t) = F^{-1} \{X(\omega)\} = \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} \frac{d\omega}{2\pi}$$

□ 주파수 변수를 ω 대신 f 를 사용하면

$$X(f) = F \{x(t)\} = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$$

$$x(t) = F^{-1} \{X(f)\} = \int_{-\infty}^{\infty} X(f) e^{j2\pi ft} df$$

□ Continuous Spectrum

- 주기 신호의 푸리에 스펙트럼이 이산 스펙트럼
- 비주기 신호의 푸리에 스펙트럼은 연속 스펙트럼

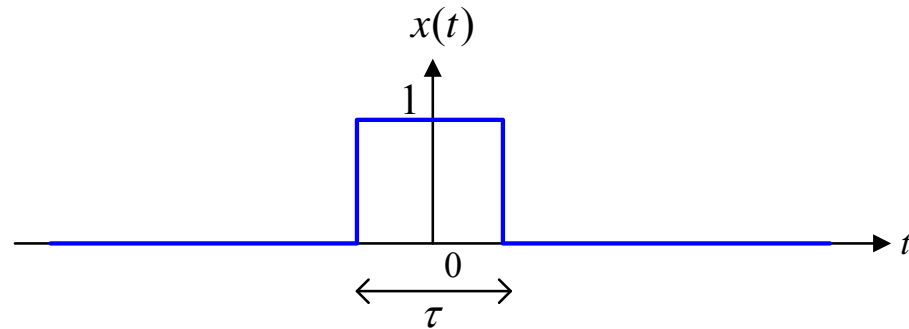
□ Amplitude spectrum and phase spectrum

$$X(\omega) = |X(\omega)| e^{j \arg X(\omega)}$$

- 1) $|X(\omega)|$: amplitude spectrum
- 2) $\arg X(\omega)$: phase spectrum

□ [예제 5.1] Rectangular Pulse with pulse width τ

$$x(t) = \Pi(t / \tau) = \begin{cases} 1 & |t| \leq \tau / 2 \\ 0 & \text{otherwise} \end{cases}$$



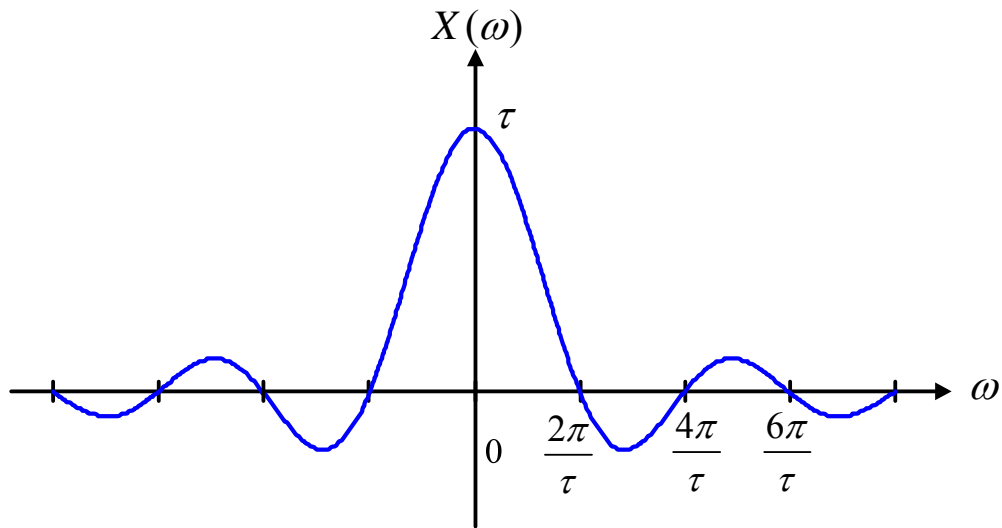
$$F \{ \Pi(t/\tau) \} = \int_{-\tau/2}^{\tau/2} e^{-j\omega t} dt$$

$$= \begin{cases} \text{If } \omega = 0, & \tau \\ \text{If } \omega \neq 0, & \left[\frac{e^{-j\omega t}}{-j\omega} \right]_{-\tau/2}^{\tau/2} = \tau \frac{\sin\left(\frac{\omega\tau}{2}\right)}{\left(\frac{\omega\tau}{2}\right)} \end{cases}$$

$$= \tau \text{Sa}\left(\frac{\omega\tau}{2}\right)$$

□ [예제 5.1] Rectangular Pulse with pulse width τ

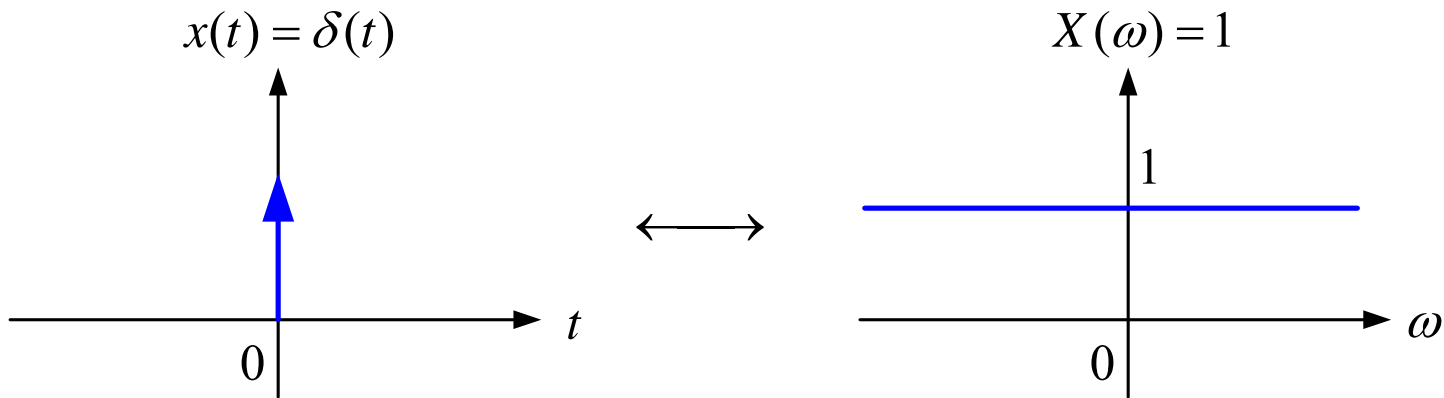
$$\Pi(t / \tau) \xleftrightarrow{F} \tau \text{Sa}\left(\frac{\omega\tau}{2}\right)$$



□ [예제 5.2] Impulse

$$F \{ \delta(t) \} = \int_{-\infty}^{\infty} \delta(t) e^{-j\omega t} dt = e^0 \int_{-\infty}^{\infty} \delta(t) dt = 1$$

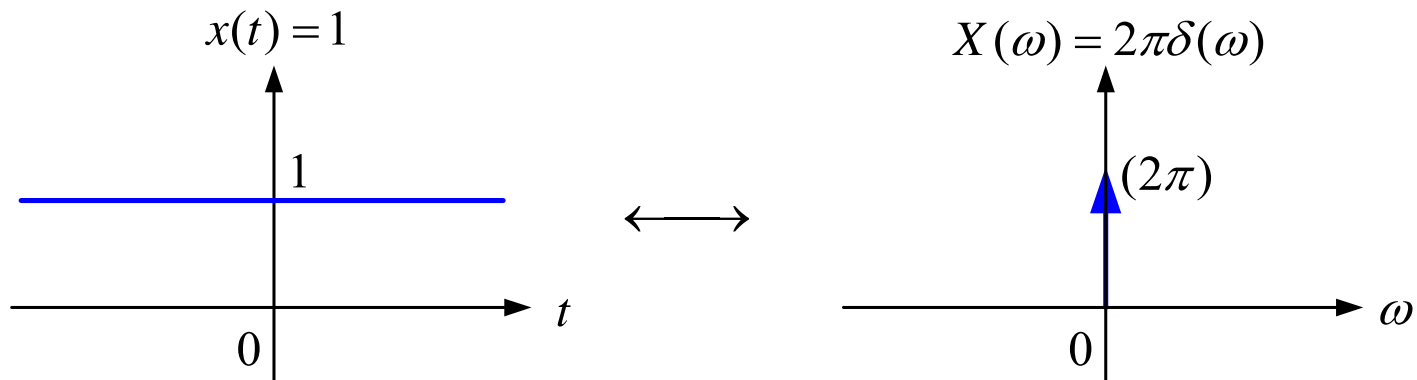
$$\delta(t) \xleftrightarrow{F} 1$$



□ [예제 5.3] Inverse Fourier transform of $\delta(\omega)$

$$F^{-1}\{\delta(\omega)\} = \int_{-\infty}^{\infty} \delta(\omega) e^{j\omega t} \frac{d\omega}{2\pi} = \int_{-\infty}^{\infty} \delta(\omega) e^0 \frac{d\omega}{2\pi} = \frac{1}{2\pi}$$

$$1 \xleftrightarrow{F} 2\pi \delta(\omega) = \delta(f)$$



□ [예제 5.4] Inverse Fourier transform of rectangular spectrum

- Rectangular spectrum of bandwidth $\tau / 2$

$$X(\omega) = \Pi(\omega / \tau) = \begin{cases} 1 & |\omega| < \tau / 2 \\ 0 & |\omega| > \tau / 2 \end{cases}$$

$$\mathbf{x}(t) = \mathbb{F}^{-1} \{ \Pi(\omega / \tau) \} = \int_{-\tau/2}^{\tau/2} e^{j\omega t} \frac{d\omega}{2\pi}$$

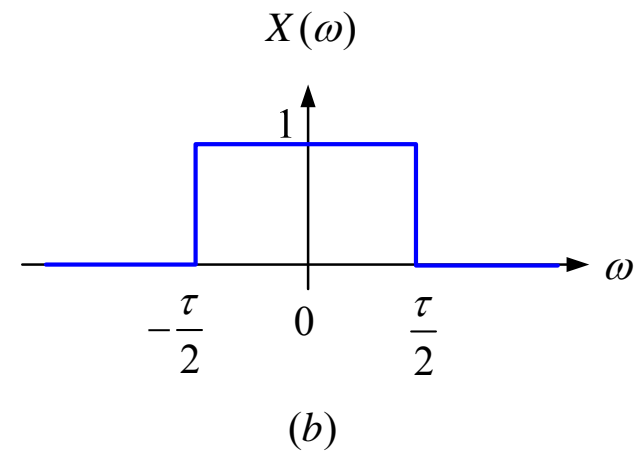
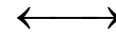
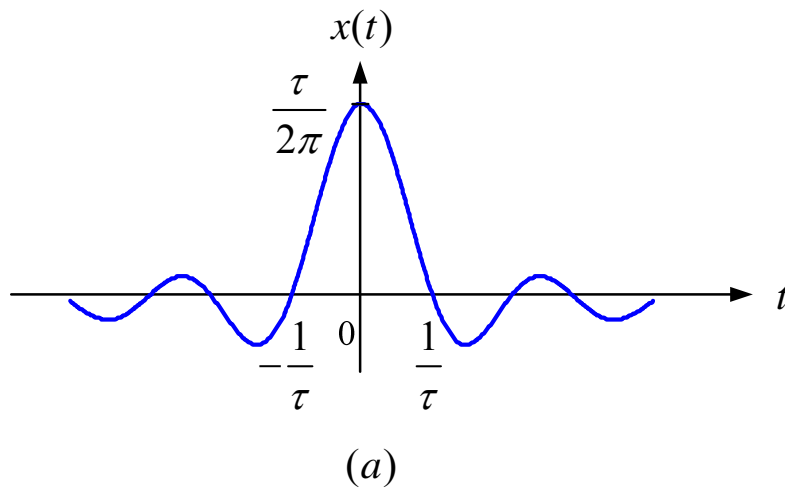
$$= \begin{cases} \text{If } t = 0, & \frac{\tau}{2\pi} \\ \text{If } t \neq 0, & \left[\frac{1}{2\pi} \frac{\exp(j\omega t)}{jt} \right]_{-\tau/2}^{\tau/2} = \frac{\tau}{2\pi} \frac{\sin\left(\frac{\tau t}{2}\right)}{\left(\frac{\tau t}{2}\right)} \end{cases}$$

$$= \frac{\tau}{2\pi} \text{Sa}\left(\frac{\tau t}{2}\right)$$

□ [예제 5.4] Inverse Fourier transform of rectangular spectrum

- Rectangular spectrum of bandwidth $\tau / 2$

$$\frac{1}{2\pi} \cdot \tau \text{Sa} \left(\frac{\tau t}{2} \right) \xleftrightarrow{\text{F}} \Pi \left(\frac{\omega}{\tau} \right)$$



□ [예제 5.5] Exponential function

$$x(t) = e^{-at} u(t), \quad a > 0$$

$$\begin{aligned} X(\omega) &= \int_{-\infty}^{\infty} e^{-at} e^{-j\omega t} u(t) dt = \int_0^{\infty} e^{-(a+j\omega)t} dt \\ &= \frac{1}{a + j\omega}, \quad a > 0 \end{aligned}$$

$$|X(\omega)| = \frac{1}{\sqrt{a^2 + \omega^2}} = \frac{1}{a} \cdot \frac{1}{\sqrt{1 + \left(\frac{\omega}{a}\right)^2}}$$

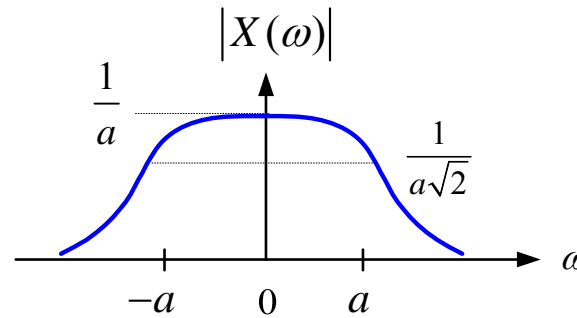
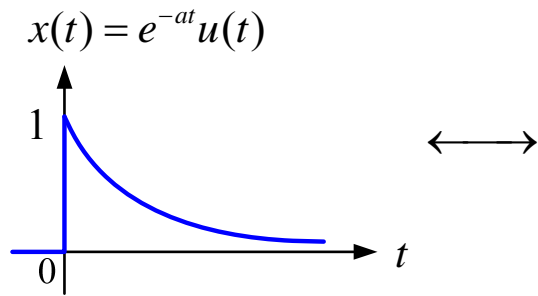
$$\arg X(\omega) = -\tan^{-1}\left(\frac{\omega}{a}\right)$$

$$|X(a)| = |X(0)| \cdot \frac{1}{\sqrt{2}}$$

$$\arg X(a) = -\frac{\pi}{4}$$

□ [예제 5.5] Exponential function

$$e^{-at}u(t) \xleftrightarrow{F} \frac{1}{a + j\omega}$$



$$|X(\omega)| = \frac{1}{a} \frac{1}{\sqrt{1 + \left(\frac{\omega}{a}\right)^2}}$$

