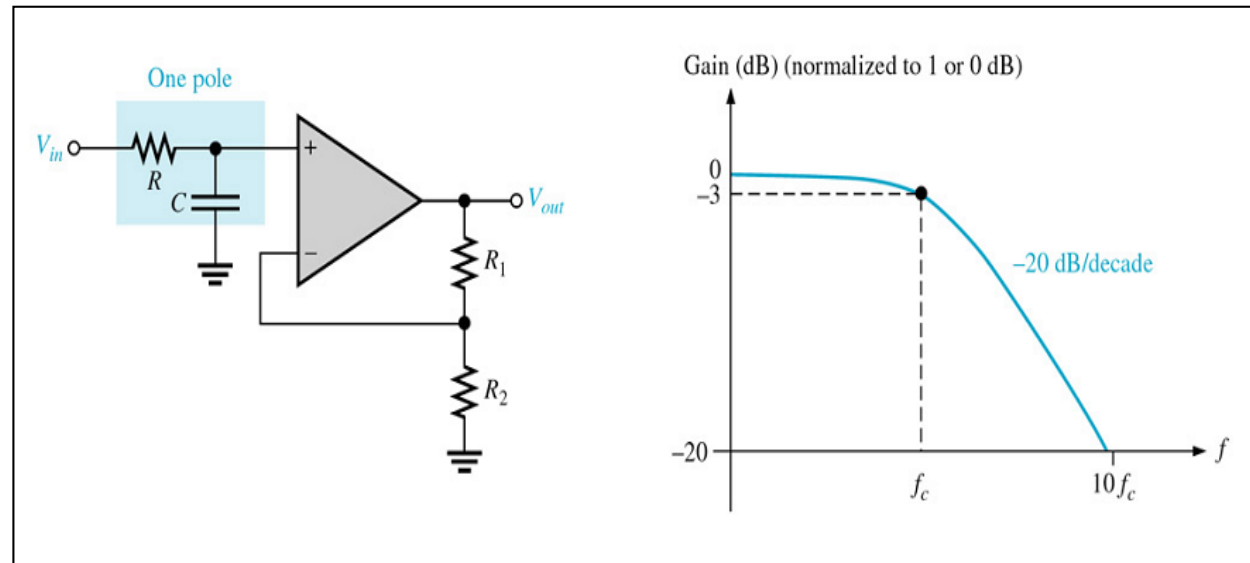


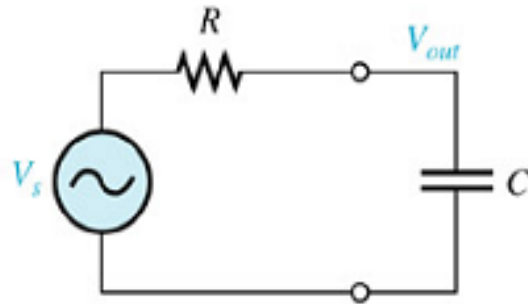
# Ch. 15 Active Filters



# 기본적인 필터 응답(Basic filter response)

Yun SeopYu

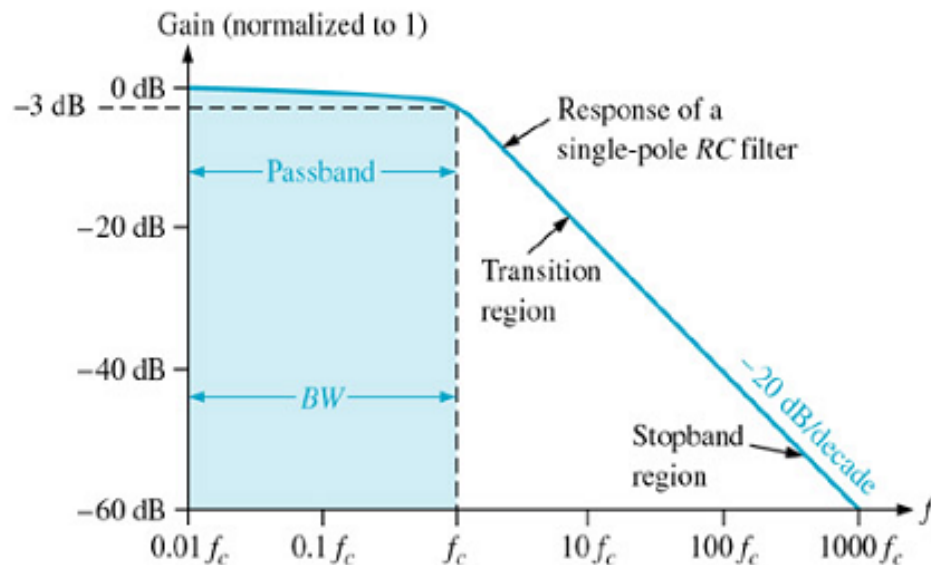
## 저역통과 필터 응답 (low-pass filter (LPF) response)



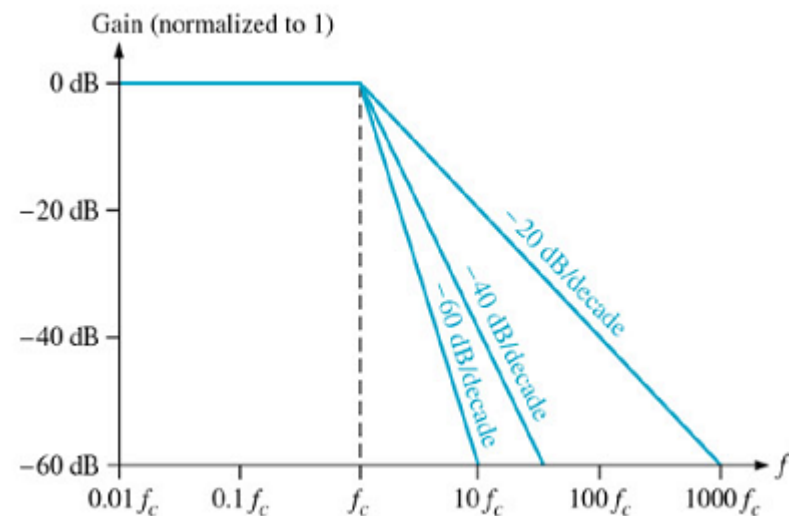
$$A_{v(db)} = 20 \log \frac{V_{out}}{V_s} = 20 \log \frac{X_C}{\sqrt{R^2 + X_C^2}} = 20 \log \frac{1}{\sqrt{1 + \frac{R^2}{X_C^2}}} = 20 \log \frac{1}{\sqrt{1 + (2\pi fRC)^2}}$$

when  $R = X_C = \frac{1}{2\pi fC} \rightarrow f_c = \frac{1}{2\pi RC}$

### Basic LPF response

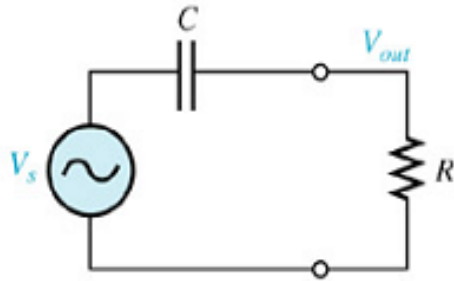


### LPF with different roll-off rates



# 기본적인 필터 응답

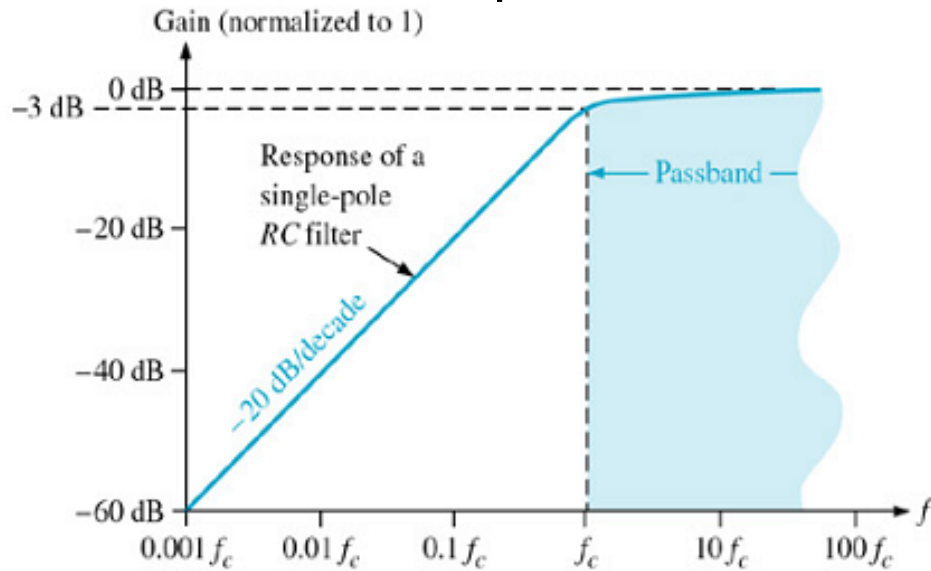
## 고역통과 필터 응답 (high-pass filter (HPF) response)



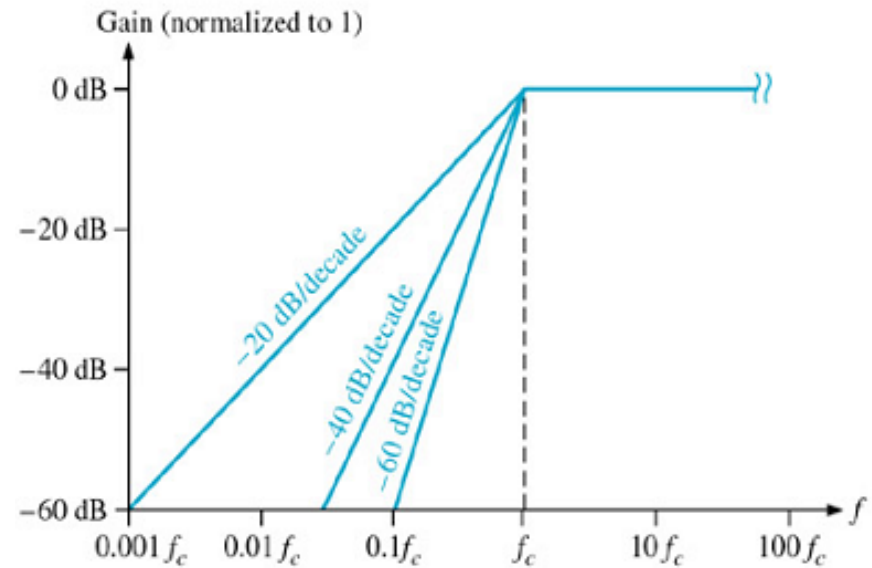
$$A_{v(db)} = 20 \log \frac{V_{out}}{V_s} = 20 \log \frac{R}{\sqrt{R^2 + X_C^2}} = 20 \log \frac{1}{\sqrt{1 + \frac{X_C^2}{R^2}}} = 20 \log \frac{1}{\sqrt{1 + \left(\frac{1}{2\pi fRC}\right)^2}}$$

when  $R = X_C = \frac{1}{2\pi fC} \rightarrow f_c = \frac{1}{2\pi RC}$

### Basic HPF response

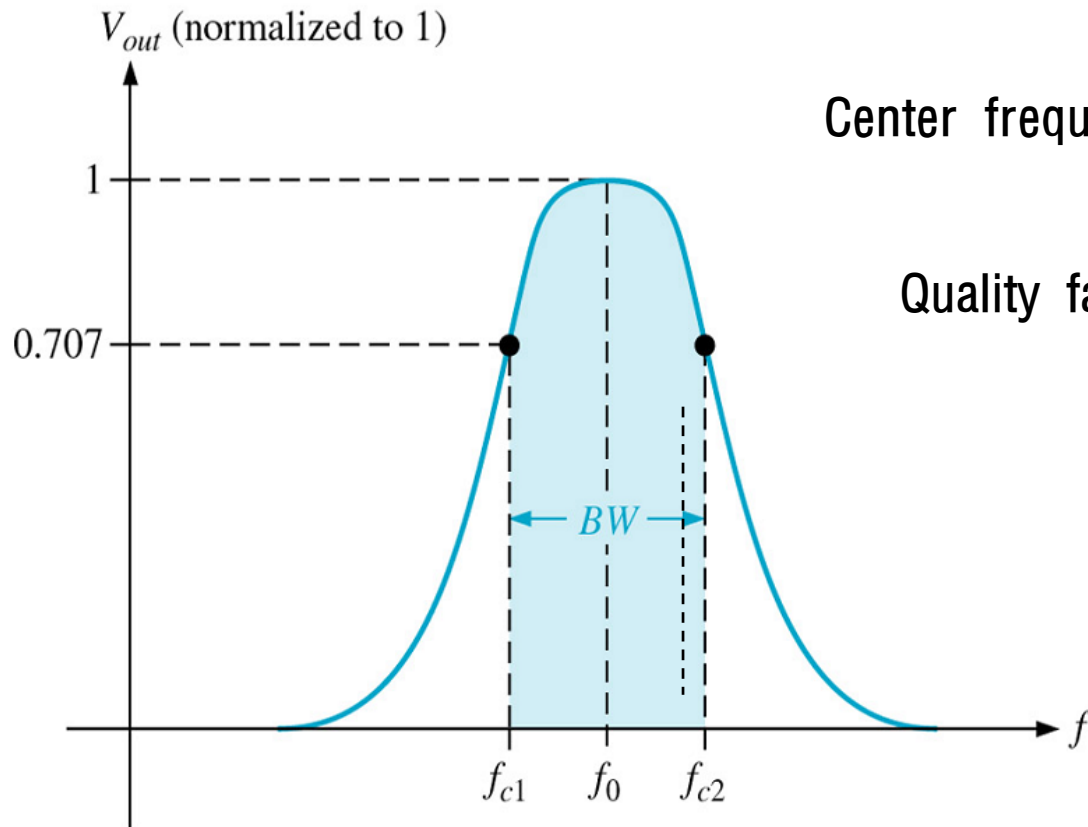


### HPF with different roll-off rates



# 기본적인 필터 응답

## 대역통과 필터 응답 (band-pass filter (BPF) response)



Center frequency:

$$f_o = \sqrt{f_{c1} f_{c2}}$$

Quality factor:

$$Q = \frac{f_o}{BW}$$

Q is an indication of the selectivity of a BPF.

Narrow BPF:  $Q > 10$ .

Wide-band BPF:  $Q < 10$ .

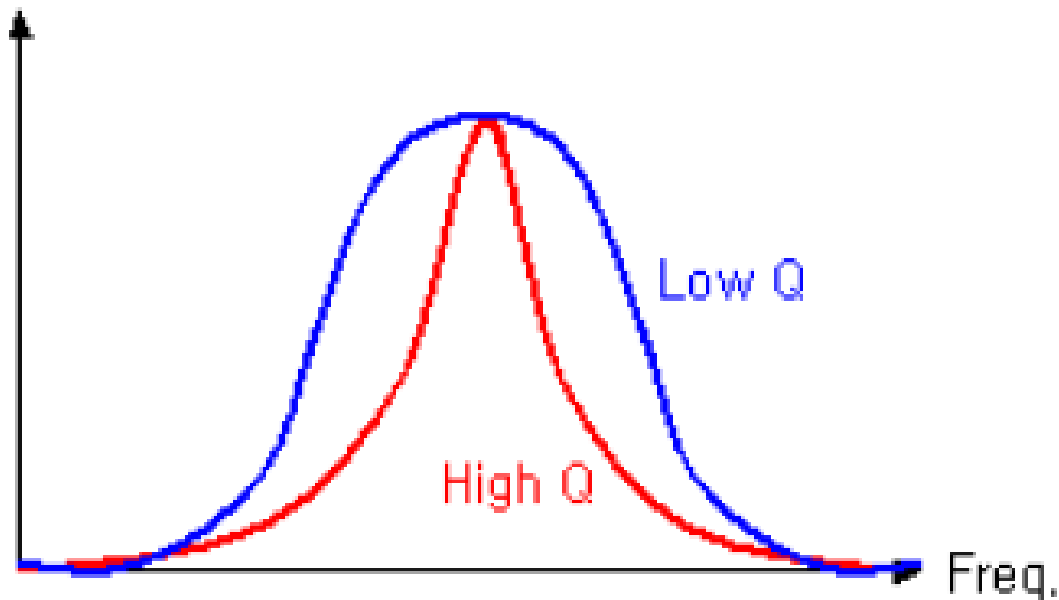
$$BW = f_{c2} - f_{c1}$$

Damping Factor:

$$DF = \frac{1}{Q}$$

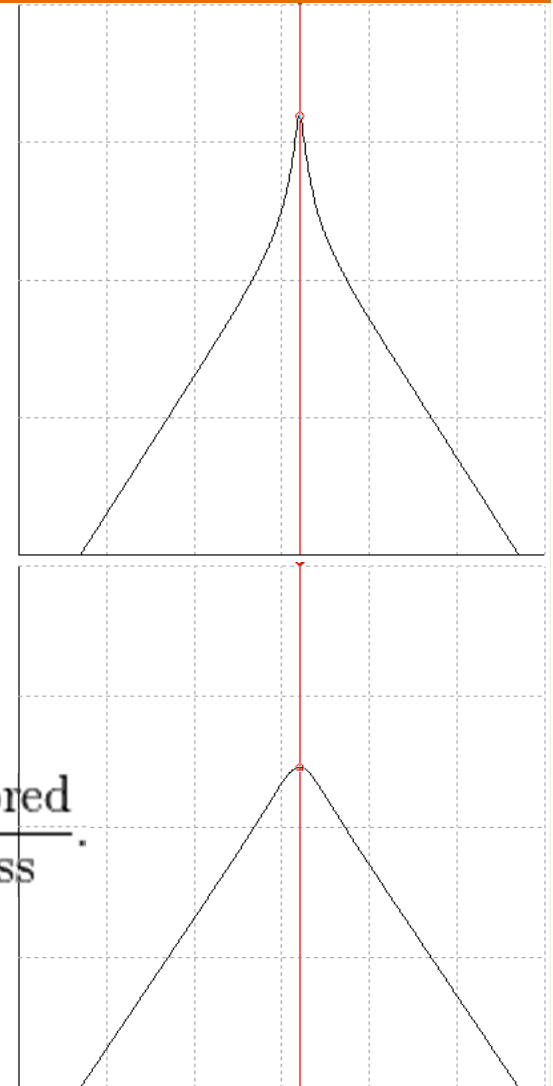
# 기본적인 필터 응답

Q 값: 주파수 선택 특성 품질



$$Q = 2\pi \times \frac{\text{Energy Stored}}{\text{Energy dissipated per cycle}} = 2\pi f_r \times \frac{\text{Energy Stored}}{\text{Power Loss}}$$

$$Q(\omega) = \omega \times \frac{\text{Maximum Energy Stored}}{\text{Power Loss}}$$

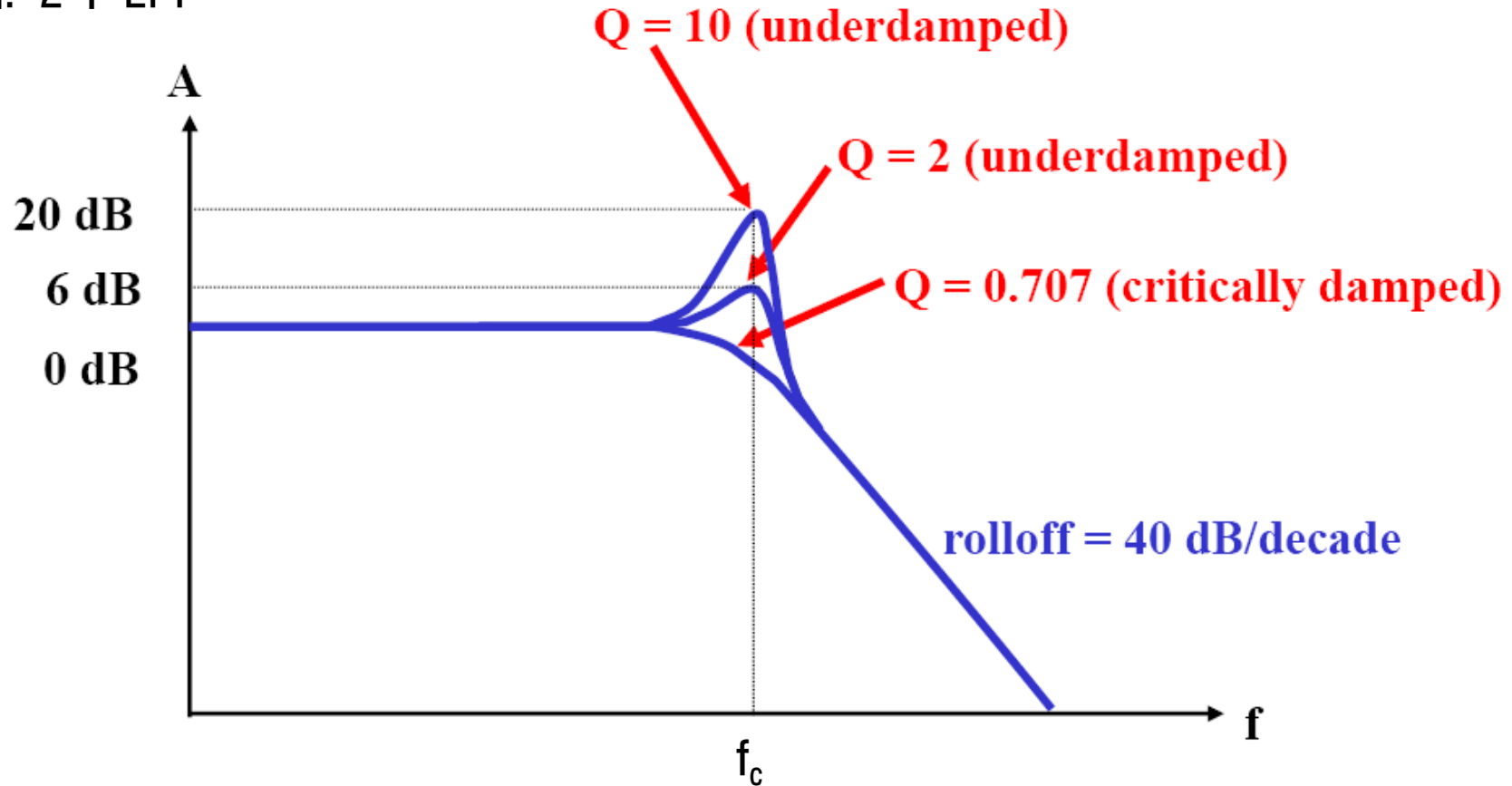


# 기본적인 필터 응답

Yun SeopYu

## Q 값: 댐핑 성질 결정

예: 2차 LPF



$$DF = \frac{1}{Q}$$

# 예제 15-1

Yun SeopYu

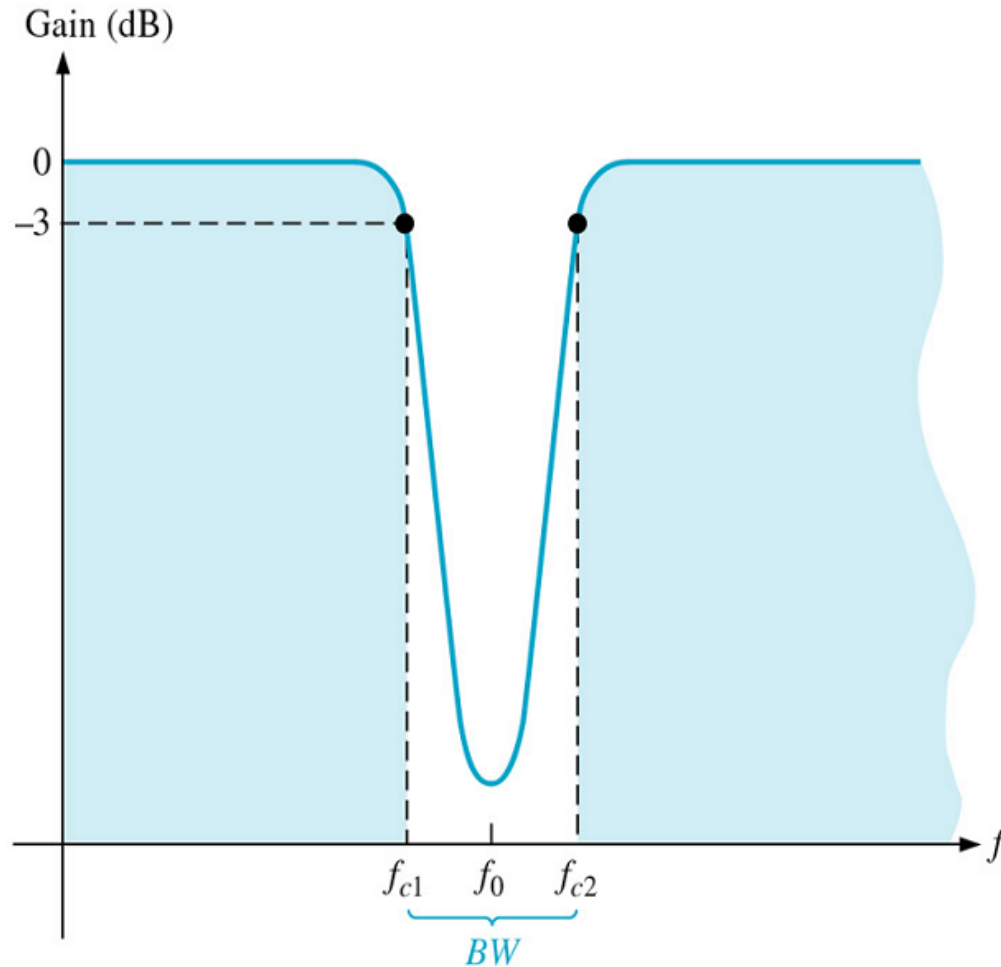
✪  $f_0 = 15 \text{ kHz}$ ,  $BW = 1 \text{ kHz}$ ,  $Q$  ?

$$Q = \frac{f_o}{BW} = \frac{15\text{k}}{1\text{k}} = 15 > 10 \rightarrow \text{협대역 (narrow BW)}$$

# 기본적인 필터 응답

Yun SeopYu

## 대역저지 필터 응답 (band-stop filter response)



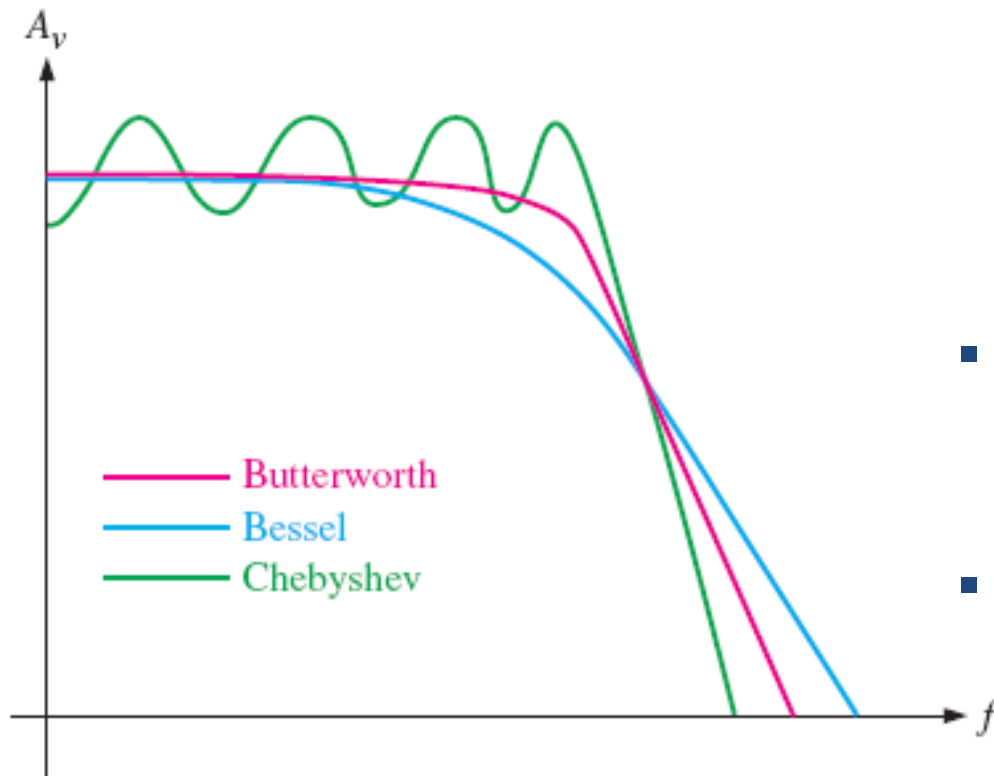
- Band-reject, or notch filter.
- 어떤 특정 주파수 대역만 저지하고 그 외 주파수 대역은 통과시키는 필터



# 필터 응답 특성(Filter response characteristics)

Yun SeopYu

## 기본 필터 응답 특성

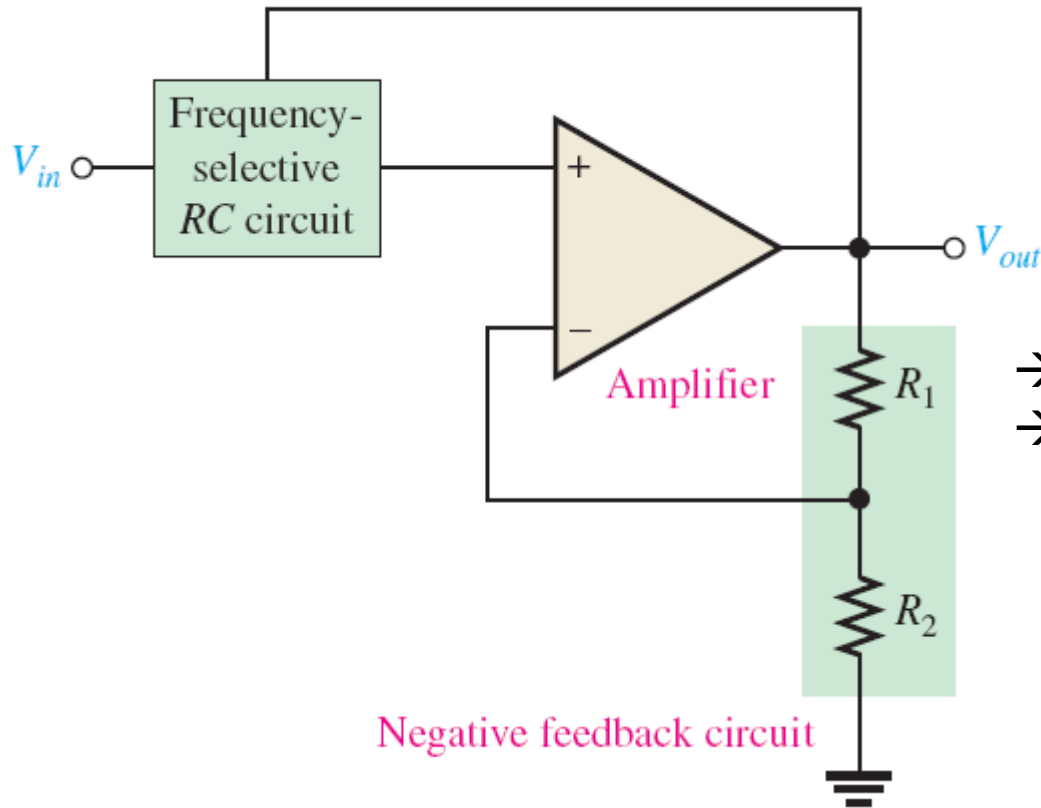


- **Butterworth:** 통과 대역 매우 평탄
  - roll-off rate:  $-20$  dB/dec/pole
  - 위상지연: 비선형
    - 펄스의 상승, 하강 edge의 주파수 부품들의 서로 다른 위상지연
    - 출력의 overshoot
  - 통과 대역의 모든 주파수의 이득이 모두 같아야 할 때
- **Chebyshev:**
  - roll-off rate  $> -20$  dB/dec/pole
  - 통과대역 리플 → 비선형 위상 응답
- **Bessel:** 선형적인 위상 응답
  - No overshoot 출력
  - roll-off rate  $< -20$  dB/dec/pole.

# 필터 응답 특성

## ❁ 댐핑 계수 (damping factor: DF)

### General diagram of active filter



DF: 필터의 응답 특성을 결정

$$DF = 2 - \frac{R_1}{R_2}$$

- 부귀환 동작으로 필터응답에 영향
- 필터의 차수 (극점 수)와 관련된 올바른 응답 특성을 구현하기 위해서 댐핑계수 값이 요구
- ← 표 15-1 참고

# 표 15-1

## Values for the Butterworth response

ORDER	ROLL-OFF DB/DECADE	1ST STAGE			2ND STAGE			3RD STAGE		
		POLES	DF	$R_1/R_2$	POLES	DF	$R_3/R_4$	POLES	DF	$R_5/R_6$
1	-20	1	Optional							
2	-40	2	1.414	0.586						
3	-60	2	1.00	1	1	1.00	1			
4	-80	2	1.848	0.152	2	0.765	1.235			
5	-100	2	1.00	1	2	1.618	0.382	1	0.618	1.382
6	-120	2	1.932	0.068	2	1.414	0.586	2	0.518	1.482

## ❁ 댐핑 계수 (damping factor: DF)

❁ 극점 (pole): 한 개의 저항과 캐패시터로 구성된 간단한 회로

- Roll-off: -20dB/decade/pole

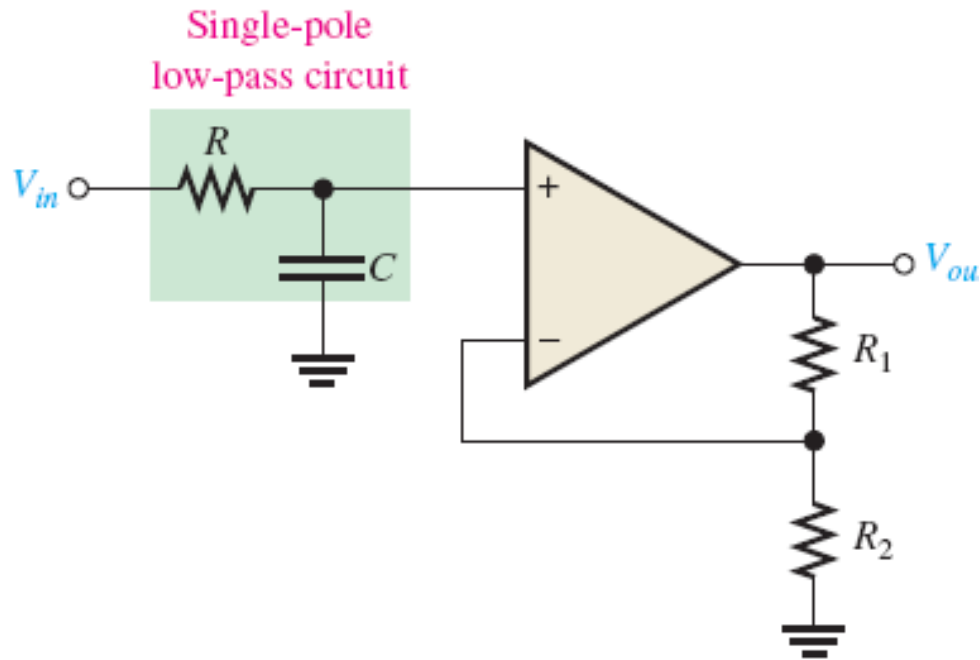
## ❁ 예제: 2차 버터워스 (Butterworth) 응답

- 극점: 2개  $\rightarrow$   $DF = 1.414$   $\leftarrow$  표15-1
- $R_1/R_2 = 2 - DF = 2 - 1.414 = 0.586$
- 비반전 증폭기 전압이득  $A_{cl(NI)} = 1 + R_1/R_2 = 1.586$
- 만약  $R_2 = 10 \text{ k}\Omega \rightarrow R_1 = 5.86 \text{ k}\Omega$

# 필터 응답 특성

Yun SeopYu

## ❁ 차단 주파수와 기울기



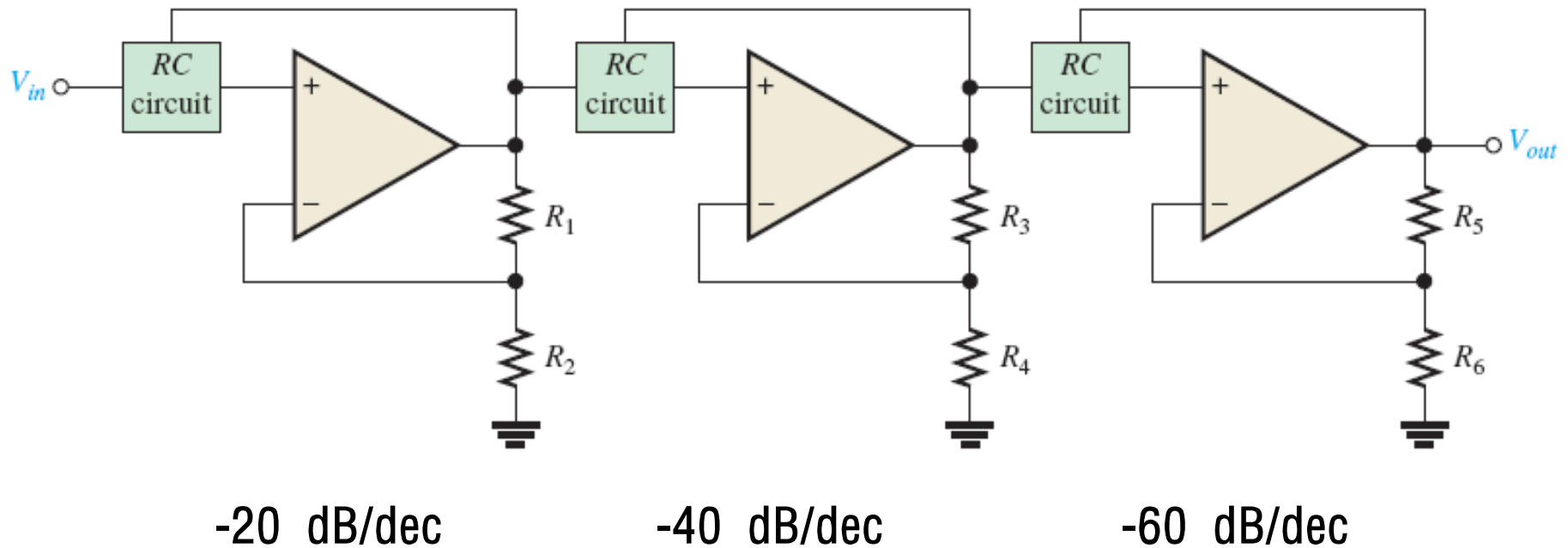
$$f_c = \frac{1}{2\pi RC}$$

Roll-off: -20 dB/dec

# 필터 응답 특성

Yun SeopYu

## 차단 주파수와 기울기



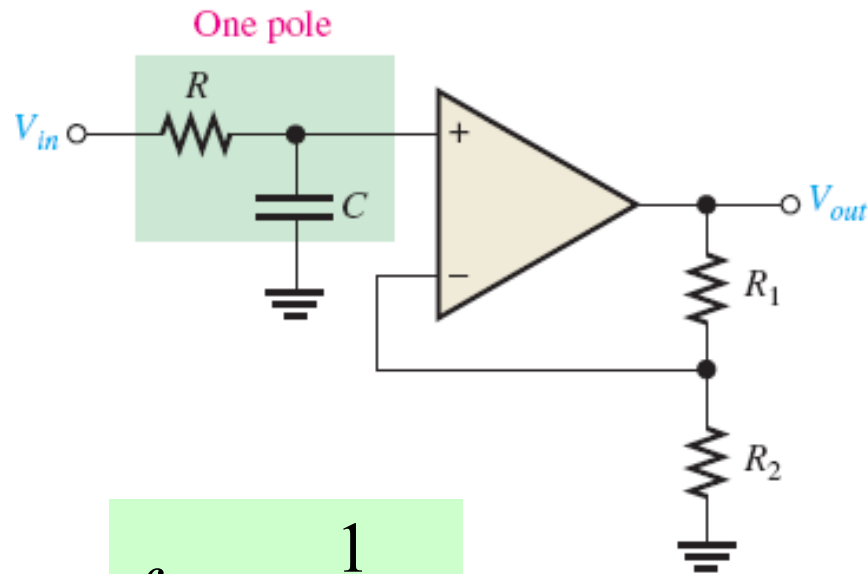
## 능동 필터의 장점

- 연산 증폭기는 필터를 통과에 의한 이득 감소를 고려하여 높은 전압 이득 제공
- 높은 입력 임피던스: 구동원의 과부하를 막아줌
- 낮은 출력 임피던스: 필터가 구동될 때 부하의 영향으로부터 필터를 보호
- 요구되는 응답을 수정할 필요 없이 넓은 주파수 범위에서 조정하기 쉬움

## 단점

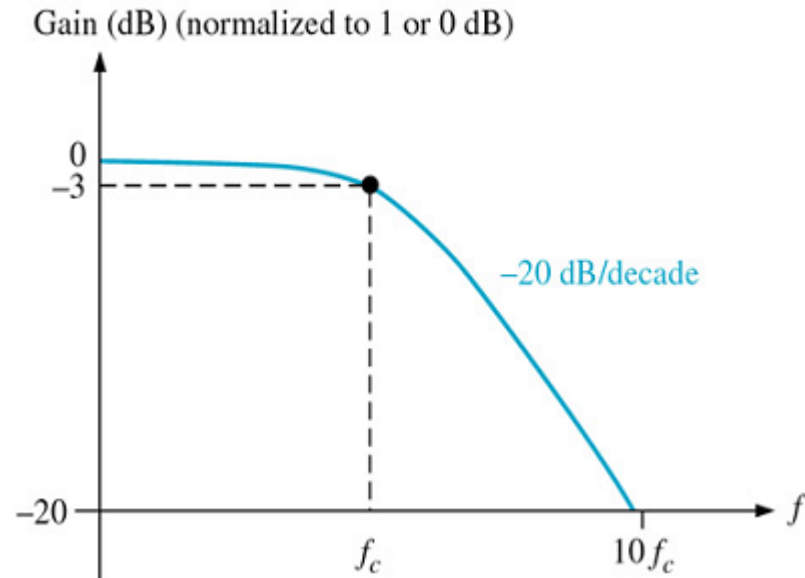
- dc 전원 필요
- 연산증폭기의 주파수 응답에 의해서 제한되어질 수 있다.

## 1차 단극 필터



$$f_c = \frac{1}{2\pi RC}$$

$$A_{cl} = 1 + \frac{R_1}{R_2}$$



Roll-off rate for a single-pole Filter: -20 dB/decade.

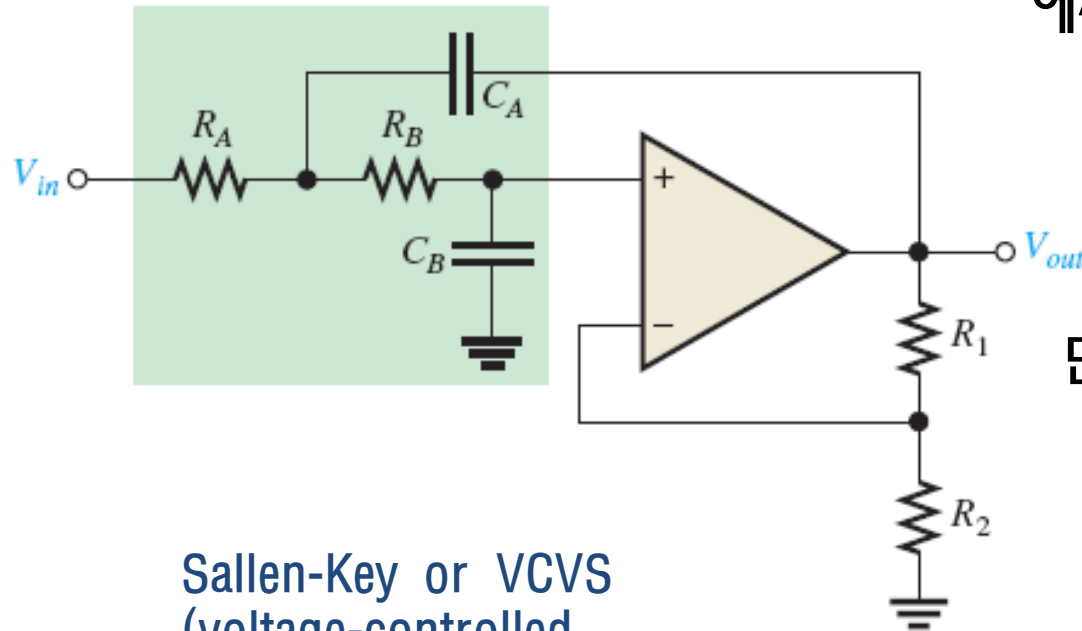
$A_{cl}$ : selectable since DF is optional for single-pole LPF



## ☉ Sallen-Key 저역통과 필터

- 40dB/decade의 기울기를 가지는 2개의 저역통과 필터
- $C_A$ 에 의해서 통과대역 모서리 근처에서 예리한 응답을 갖는다

Two-pole low-pass circuit



Sallen-Key or VCVS (voltage-controlled voltage-source) second-order low-pass filter

$$f_c = \frac{1}{2\pi \sqrt{R_A R_B C_A C_B}}$$

만약  $R_A = R_B = R, C_A = C_B = C$

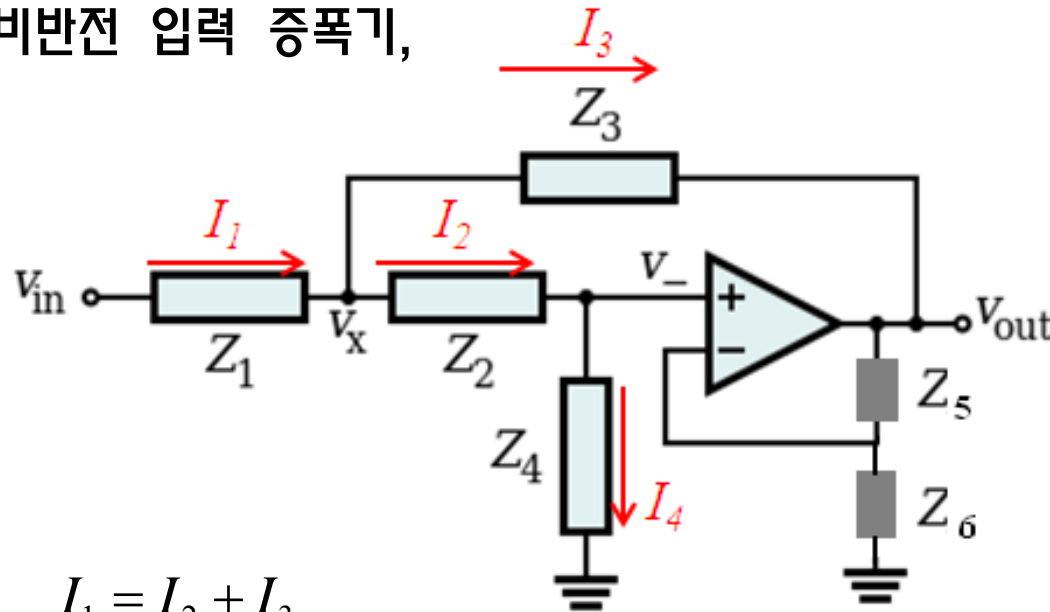
$$f_c = \frac{1}{2\pi RC}$$

For a Butterworth 2<sup>nd</sup>-order response, DF = 1.414; therefore,  $R_1/R_2 = 0.586$ .

# Sallen-Key 일반 필터 회로

Yun SeopYu

⊕ 비반전 입력 증폭기,



$$I_+ \approx I_- \approx 0$$

$$V_+ \approx V_- = \frac{Z_6}{Z_5 + Z_6} V_{out} = BV_{out}$$

$$I_1 = I_2 + I_3$$

$$\frac{V_{in} - V_x}{Z_1} = \frac{V_x - V_-}{Z_2} + \frac{V_x - V_{out}}{Z_3} \quad \text{-(1)}$$

$$I_2 = I_4$$

$$\frac{V_x - V_-}{Z_2} = \frac{V_-}{Z_4} \quad \longrightarrow \quad V_x = \left(1 + \frac{Z_2}{Z_4}\right) V_- = \left(1 + \frac{Z_2}{Z_4}\right) BV_{out} \quad \text{-(2)}$$

# Sallen-Key 일반 필터 회로

## ⊕ 비반전 입력 증폭기,

- 식 (2)를 식 (1)에 대입하여 전개하면,

$$\frac{V_{in} - \left(1 + \frac{Z_2}{Z_4}\right)BV_{out}}{Z_1} = \frac{\left(1 + \frac{Z_2}{Z_4}\right)BV_{out} - BV_{out}}{Z_2} + \frac{\left(1 + \frac{Z_2}{Z_4}\right)BV_{out} - V_{out}}{Z_3}$$

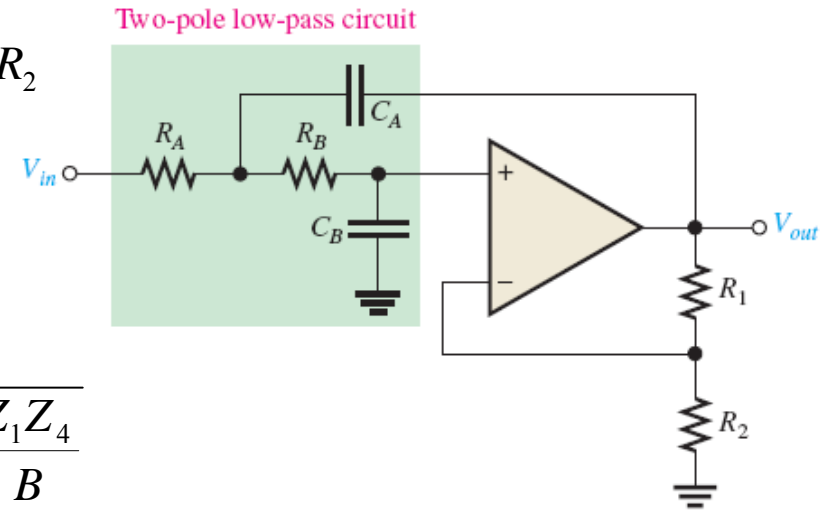
$$V_{in} - \left(1 + \frac{Z_2}{Z_4}\right)BV_{out} = \left(\frac{Z_1}{Z_4} + \frac{Z_1}{Z_3} + \frac{Z_1}{Z_3} \frac{Z_2}{Z_4} - \frac{1}{B} \frac{Z_1}{Z_3}\right)BV_{out}$$

$$A_v = \frac{V_{out}}{V_{in}} = \frac{Z_3 Z_4}{(Z_3 Z_4 + Z_1 Z_3 + Z_1 Z_4 + Z_1 Z_2 + Z_3 Z_2)B - Z_1 Z_4}$$

# Sallen-Key 저역 통과 필터

Yun SeopYu

$$Z_1 = R_A, Z_2 = R_B, Z_3 = \frac{1}{sC_A}, Z_4 = \frac{1}{sC_B}, Z_5 = R_1, Z_6 = R_2$$



$$A_v = \frac{V_{out}}{V_{in}} = \frac{\frac{Z_3 Z_4}{B}}{Z_3 Z_4 + Z_1 Z_3 + Z_1 Z_4 + Z_1 Z_2 + Z_3 Z_2 - \frac{Z_1 Z_4}{B}}$$

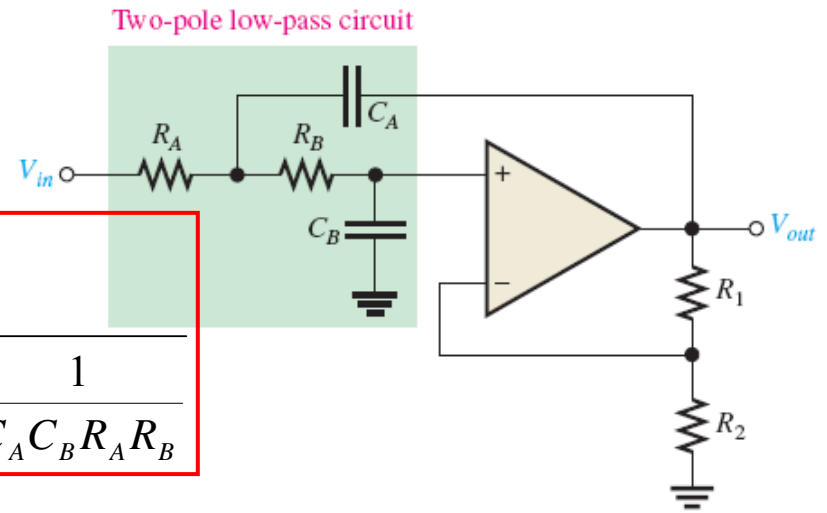
$$= \frac{\frac{1}{sC_A} \frac{1}{sC_B} \frac{1}{B}}{\frac{1}{sC_A} \frac{1}{sC_B} + \frac{R_A}{sC_A} + \frac{R_A}{sC_B} + R_A R_B + \frac{R_B}{sC_A} - \frac{R_A}{sC_B} \frac{1}{B}}$$

$$= \frac{\frac{1}{C_A C_B R_A R_B} \frac{1}{B}}{s^2 + s \left[ \frac{R_A + R_B}{C_A R_A R_B} + \frac{1}{C_B R_B} \left( 1 - \frac{1}{B} \right) \right] \frac{(R_1 + R_2)}{C_1 R_1 R_2} + \frac{1}{C_A C_B R_A R_B}}$$

# Sallen-Key 저역 통과 필터

$$Z_1 = R_A, Z_2 = R_B, Z_3 = \frac{1}{sC_A}, Z_4 = \frac{1}{sC_B}, Z_5 = R_1, Z_6 = R_2$$

$$A_v = \frac{V_{out}}{V_{in}} = \frac{\frac{1}{C_A C_B R_A R_B} \frac{1}{B}}{s^2 + s \left[ \frac{R_A + R_B}{C_A R_A R_B} + \frac{1}{C_B R_B} \left( 1 - \frac{1}{B} \right) \right] \frac{(R_1 + R_2)}{C_1 R_1 R_2} + \frac{1}{C_A C_B R_A R_B}}$$



- 2차 저역통과 필터 전달함수

$$A_v \rightarrow H(s) = \frac{A_0 \omega_0^2}{s^2 + \frac{\omega_0}{Q} s + \omega_0^2}$$

$$\omega_0^2 = (2\pi f_0)^2 = \frac{1}{C_A C_B R_A R_B}$$

$$\rightarrow f_0 = \frac{1}{2\pi \sqrt{C_A C_B R_A R_B}}$$

$$A_0 = 1/B = 1 + R_1/R_2$$

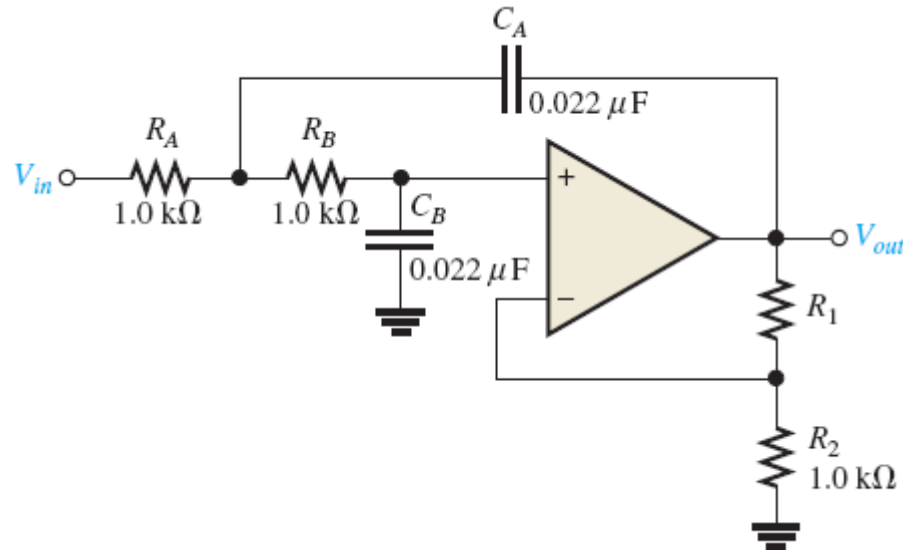
$$\frac{\omega_0}{Q} = \frac{R_A + R_B}{C_A R_A R_B} - \frac{1}{C_B R_B} \left( \frac{R_2}{R_1} \right) \rightarrow Q = \omega_0 \frac{1}{\frac{R_A + R_B}{C_A R_A R_B} - \frac{1}{C_B R_B} \left( \frac{R_2}{R_1} \right)} \rightarrow Q = \frac{1}{\frac{(R_A + R_B) C_B}{C_A R_A R_B} - C_A R_A \left( \frac{R_2}{R_1} \right)}$$

## 예제 15-3

Yun SeopYu

저역 통과 필터 → 차단주파수  $f_c$ ?,  $R_1$  (버터워스 응답)?

$$R_A = R_B = R_2 = 1 \text{ k}\Omega, C_A = C_B = 0.022 \text{ }\mu\text{F}$$



$$f_c = \frac{1}{2\pi\sqrt{R_A R_B C_A C_B}} = \frac{1}{2\pi RC} = \frac{1}{2\pi(1\text{k})(0.022\mu)} = 7.23\text{kHz}$$

2차 버터워스 → DF = 1.414

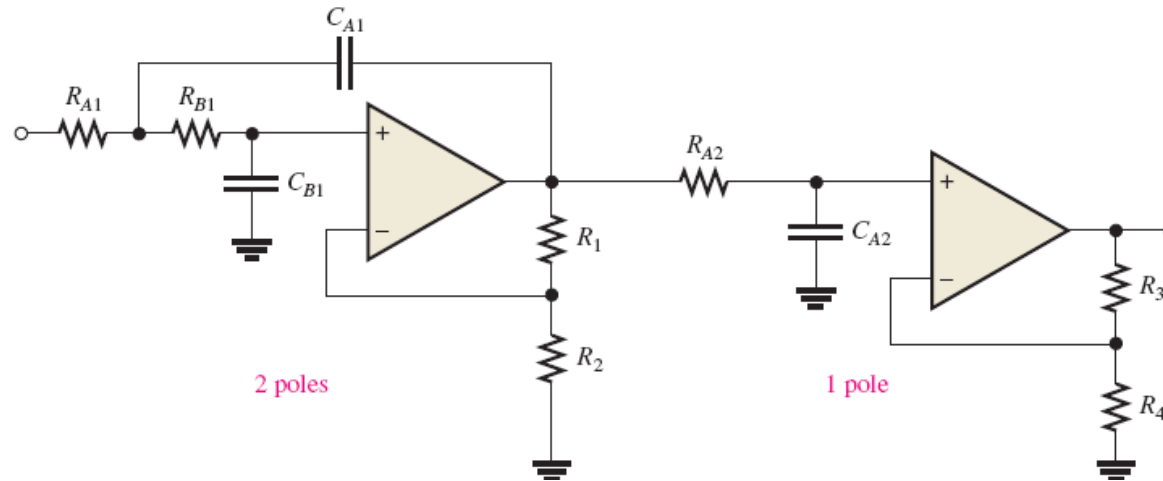
$$\rightarrow R_1/R_2 = 2 - DF = 2 - 1.414 = 0.586$$

$$\rightarrow R_1 = 0.586R_2 = (0.586)(1\text{k}) = 586 \text{ }\Omega$$

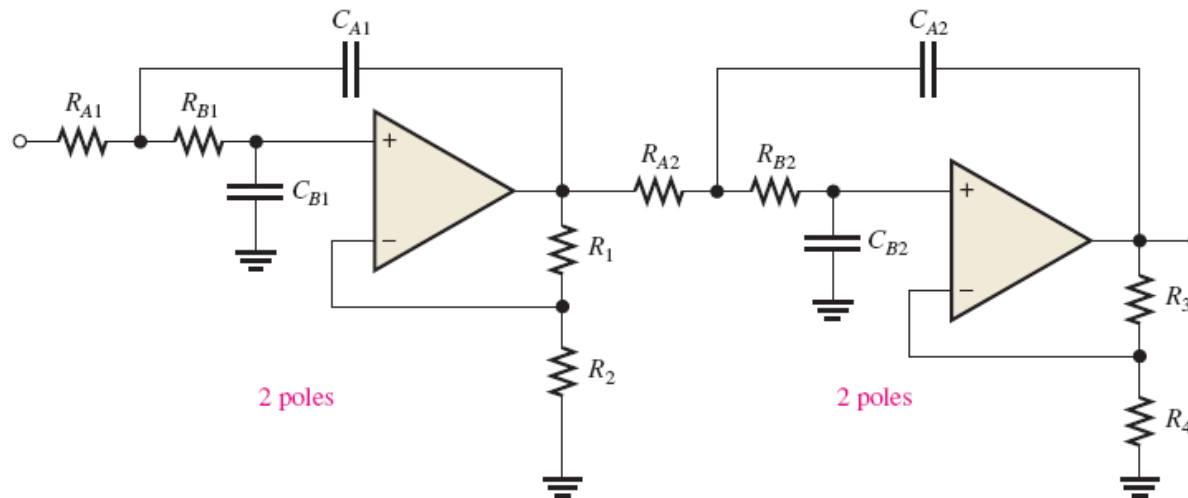
# 능동 저역 통과 필터

## Cascaded Low-Pass Filters

높은 기울기를 얻기 위해서 저역통과 필터를 종속 연결함



3극: -60 dB/dec



4극: -80 dB/dec

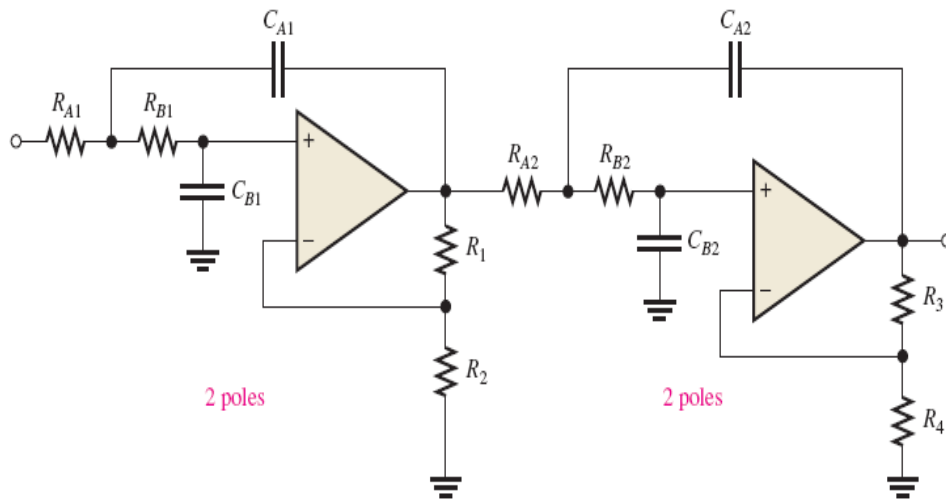
# 예제 15-4

4차 저역통과 필터 → C ?,  $R_1 = R_3$  (버터워스 응답)?

$R_{A1} = R_{B1} = R_{A2} = R_{B2} = R_2 = R_4 = 1.8 \text{ k}\Omega$

$C_{A1} = C_{B1} = C_{A2} = C_{B2} = C$

$f_c = 2680 \text{ Hz}$  라 하자



4차 버터워스:

- 첫째단:  $DF = 1.848$

→  $R_1/R_2 = 2 - DF$   
 $= 2 - 1.848 = 0.152$

→  $R_1 = 0.152R_2$   
 $= (0.152)(1.8\text{k}) = 274\Omega$

- 둘째단:  $DF = 0.765$

→  $R_3/R_4 = 2 - DF$   
 $= 2 - 0.765 = 1.235$

→  $R_3 = 1.235R_4$   
 $= (1.235)(1.8\text{k})$   
 $= 2.22\text{k}\Omega$

$f_c = \frac{1}{2\pi\sqrt{R_A R_B C_A C_B}} = \frac{1}{2\pi RC} = \frac{1}{2\pi(1.8\text{k})C} = 2.68\text{kHz}$

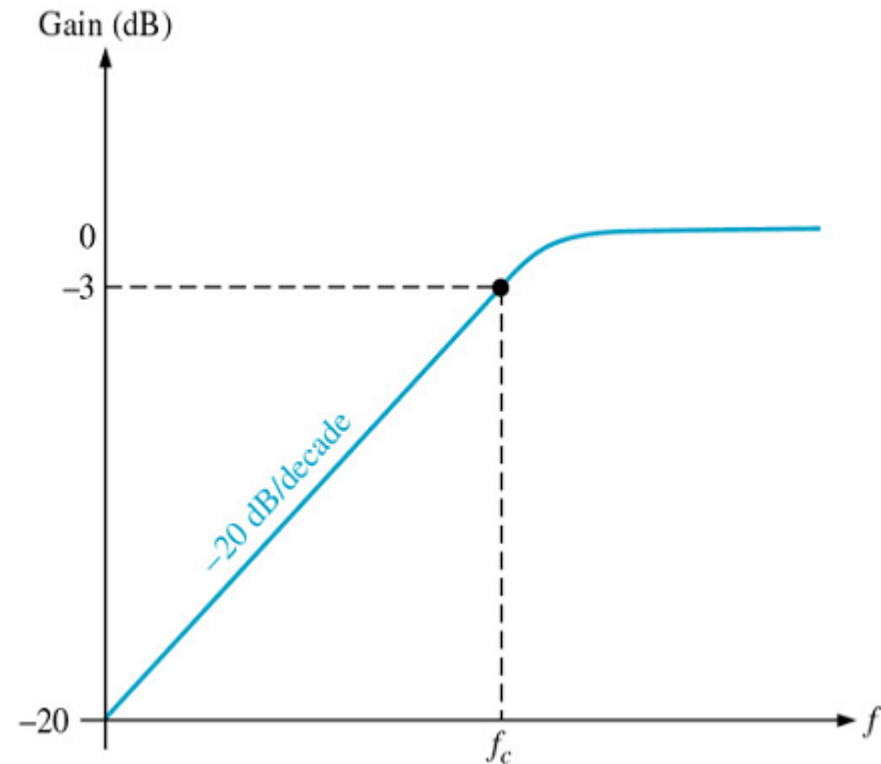
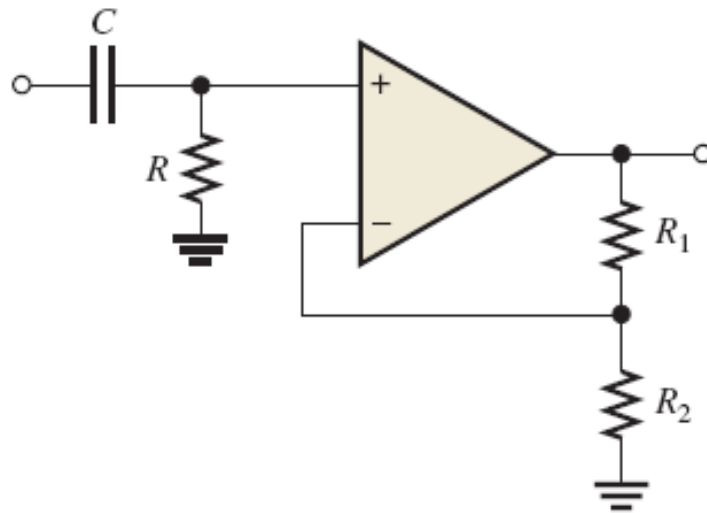
→  $C = \frac{1}{2\pi(1.8\text{k})(2.68\text{k})} = 0.033\mu\text{F}$



# 능동 고역 통과 필터 (Active High-Pass Filter)

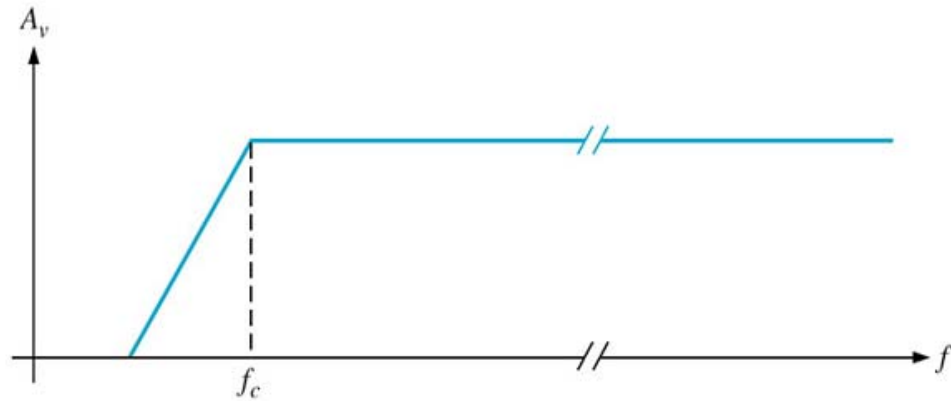
Yun SeopYu

## 1차 (단극) 필터

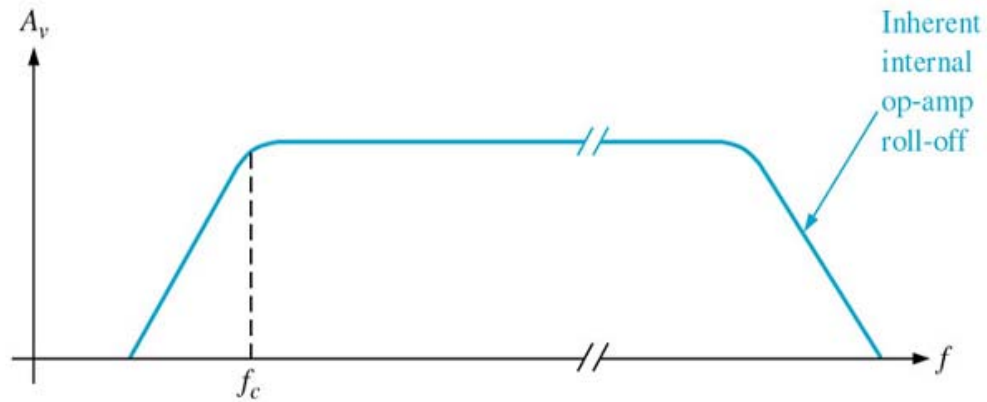


- Roll-off rate,  $f_c$ ,  $A_{cl}$  : LPF와 비슷
- 이상적인 HPF는  $f_c$  이상의 주파수는 모두 통과
- 그러나, 연산증폭기는 고주파에서 연산증폭기의 내부 RC로 인해 고주파에서 roll-off가 일어남

## 1차 (단극) 필터

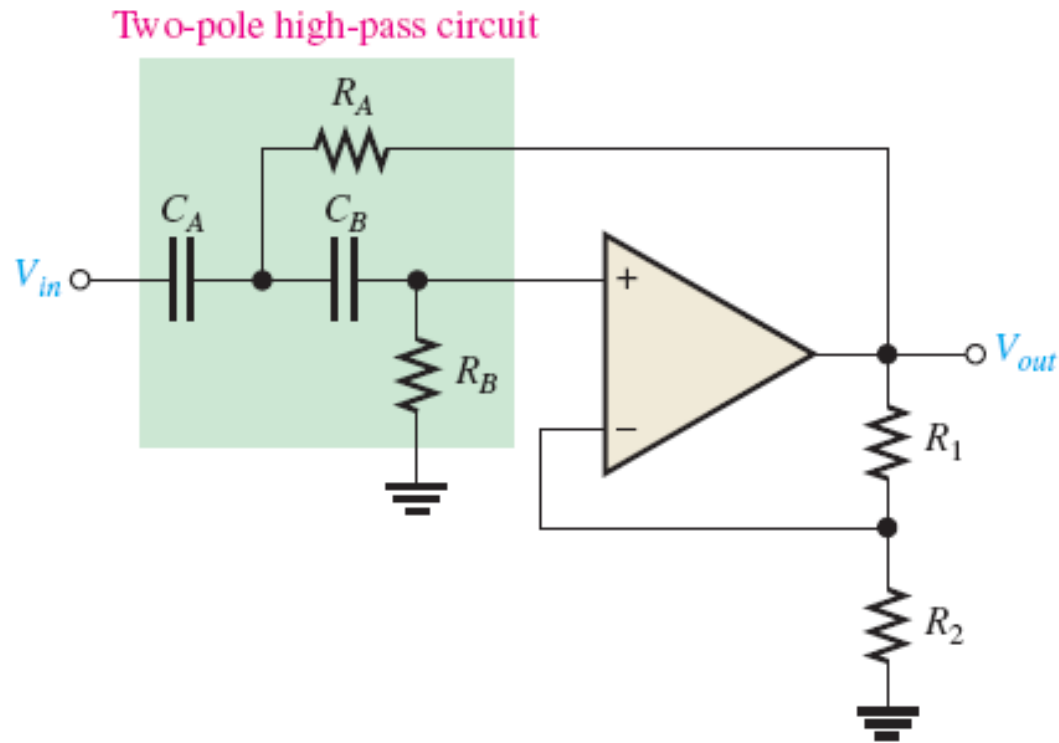


(a) Ideal



(b) Nonideal

## ❁ Sallen-Key 고역통과 필터



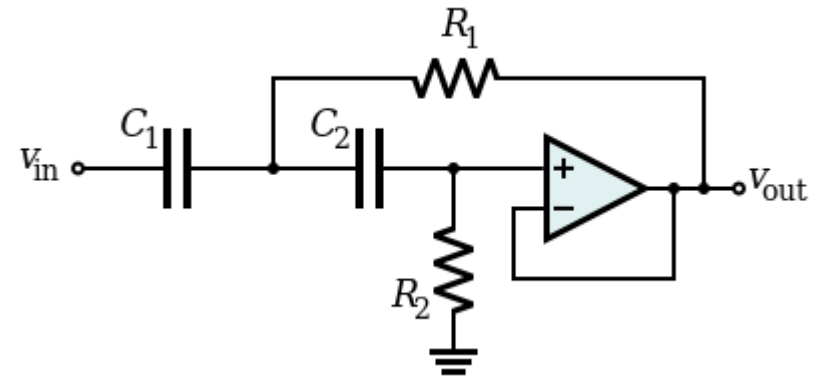
→ -40dB/decade의 기울기를 가지는 2개의 고역통과 필터

# Sallen-Key 고역 통과 필터

Yun SeopYu

$$Z_1 = \frac{1}{sC_1}, \quad Z_2 = \frac{1}{sC_2}, \quad Z_3 = R_1, \quad Z_4 = R_2$$

$$\begin{aligned} A_v &= \left| \frac{V_{out}}{V_{in}} \right| = \frac{Z_3 Z_4}{Z_1 Z_2 + Z_3 (Z_1 + Z_2) + Z_3 Z_4} \\ &= \frac{R_1 R_2}{\frac{1}{sC_1} \frac{1}{sC_2} + R_1 \left( \frac{1}{sC_1} + \frac{1}{sC_2} \right) + R_1 R_2} \\ &= \frac{sC_1 sC_2 R_1 R_2}{1 + R_1 (sC_1 + sC_2) + sC_1 sC_2 R_1 R_2} \\ &= \frac{s^2}{s^2 + s \frac{(C_1 + C_2)}{C_1 C_2 R_2} + \frac{1}{C_1 C_2 R_1 R_2}} \end{aligned}$$



- 2차 고역통과 필터 전달함수

$$A_v \rightarrow H(s) = \frac{A_0 s^2}{s^2 + \frac{\omega_0}{Q} s + \omega_0^2}$$

$$A_0 = 1, \quad \omega_0^2 = (2\pi f_0)^2 = \frac{1}{C_1 C_2 R_1 R_2}$$

$$\rightarrow f_0 = \frac{1}{2\pi \sqrt{C_1 C_2 R_1 R_2}}$$

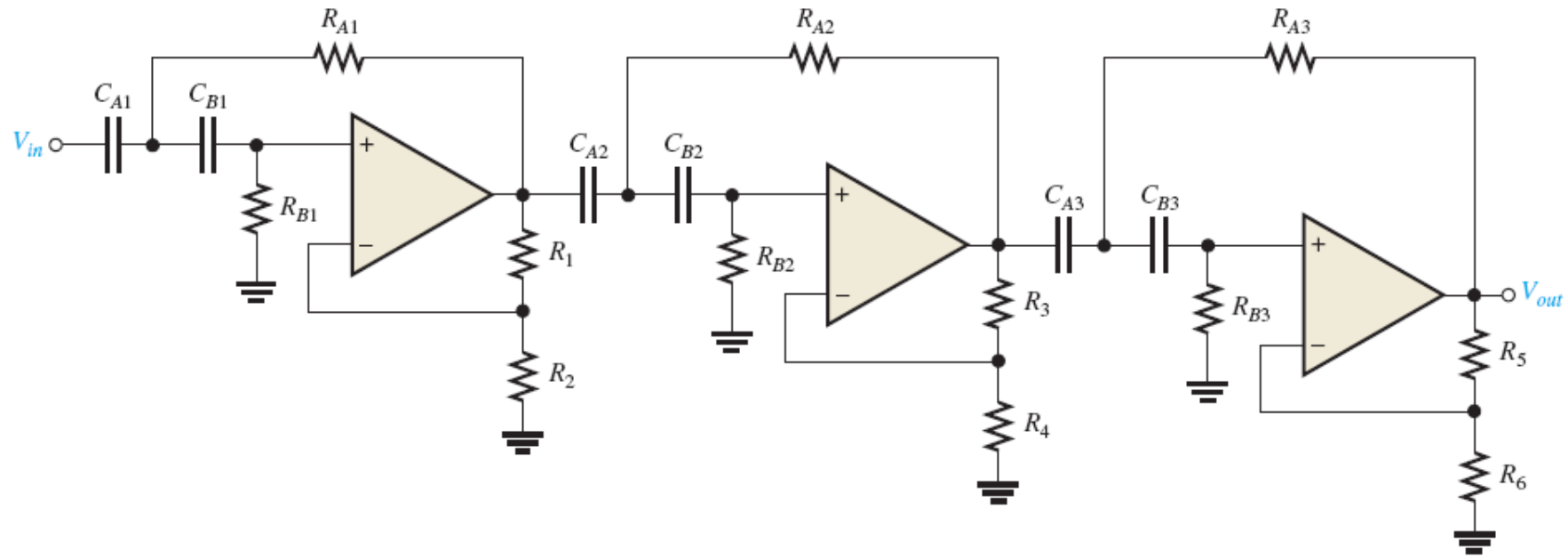
$$\frac{\omega_0}{Q} = \frac{C_1 + C_2}{C_1 C_2 R_2} = \frac{1}{R_2} \left( \frac{1}{C_1} + \frac{1}{C_2} \right)$$

$$Q = \omega_0 \frac{C_1 C_2 R_2}{C_1 + C_2} = \frac{\sqrt{C_1 C_2 R_1 R_2}}{R_1 (C_1 + C_2)}$$

# 능동 고역 통과 필터

Yun SeopYu

- ⊕ 종속접속한 고역통과 필터: roll-off 기울기를 증가



6극: -120 dB/dec

# 예제 15-5

## Shallen-key 고역통과 필터

- ❖  $f_c = 10 \text{ kHz}$ , 2차 버터워스 응답
- ❖ 가정)  $R = R_A = R_B = R_2 = 3.3 \text{ k}\Omega$
- ❖  $C = C_A = C_B$ 라 하자

$$f_c = \frac{1}{2\pi RC} = \frac{1}{2\pi(3.3\text{k})C} = 10\text{kHz}$$

$$\rightarrow C = \frac{1}{2\pi(3.3\text{k})(10\text{k})} = 0.0048\mu\text{F}$$

2차 버터워스  $\rightarrow DF = 1.414$

$$\rightarrow R_1/R_2 = 2 - DF = 2 - 1.414 = 0.586$$

$$\rightarrow R_1 = 0.586R_2 = (0.586)(3.3\text{k}) = 1.93 \text{ k}\Omega$$

